

# STM32L011x3 STM32L011x4

# Access line ultra-low-power 32-bit MCU Arm<sup>®</sup>-based Cortex<sup>®</sup>-M0+, up to 16KB Flash, 2KB SRAM, 512B EEPROM, ADC

# **Features**

- Ultra-low-power platform
  - 1.65 V to 3.6 V power supply
  - -40 to 125 °C temperature range
  - 0.23 µA Standby mode (2 wakeup pins)
  - 0.29 µA Stop mode (16 wakeup lines)
  - 0.54 µA Stop mode + RTC + 2 KB RAM
  - retention Down to 76 µA/MHz in Run mode

  - 5 µs wakeup time (from Flash memory) 41 µA 12-bit ADC conversion at 10 ksps
- Core: Arm<sup>®</sup> 32-bit Cortex<sup>®</sup>-M0+
- From 32 kHz to 32 MHz max.
- 0.95 DMIPS/MHz
- Reset and supply management
- Ultra-safe, low-power BOR (brownout reset) with 5 selectable thresholds
- Ultralow power POR/PDR
- Programmable voltage detector (PVD)
- Clock sources
  - 0 to 32 MHz external clock
  - 32 kHz oscillator for RTC with calibration
  - High speed internal 16 MHz factory-trimmed RC (+/-1%)
  - Internal low-power 37 kHz RC
  - Internal multispeed low-power 65 kHz to 4.2 MHz RC
  - PLL for CPU clock
- Pre-programmed bootloader
  - USART, SPI supported
- Development support
  - Serial wire debug supported
- Up to 28 fast I/Os (23 I/Os 5V tolerant)
- Memories
  - Up to 16 KB Flash memory with ECC
  - 2 KB RAM
  - 512 B of data EEPROM with ECC
  - 20-byte backup register

This is information on a product in full production.

Sector protection against R/W operation



UFQFPN32 5x5 mm



Datasheet - production data

WLCSP25 2.133x2.070 mm

- Rich Analog peripherals
  - 12-bit ADC 1.14 Msps up to 10 channels (down to 1.65 V)
  - 2x ultra-low-power comparators (window mode and wake up capability, down to 1.65 V)
- 5-channel DMA controller, supporting ADC, SPI, I2C, USART, Timers
- 4x peripherals communication interface
- 1x USART (ISO 7816, IrDA), 1x UART (low power)
- 1x SPI 16 Mbits/s
- 1x I2C (SMBus/PMBus)
- 7x timers: 1x 16-bit with up to 4 channels, 1x 16-bit with up to 2 channels, 1x 16-bit ultra-low-power timer, 1x SysTick, 1x RTC and 2x watchdogs (independent/window)
- CRC calculation unit, 96-bit unique ID
- All packages are ECOPACK<sup>®</sup>2

#### Table 1. Device summary

Reference	Part number
STM32L011x3	STM32L011G3, STM32L011K3, STM32L011E3, STM32L011F3, STM32L011D3
STM32L011x4	STM32L011G4, STM32L011K4, STM32L011E4, STM32L011F4, STM32L011D4

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# 1 Introduction

The ultra-low-power STM32L011x3/4 family includes devices in 7 different package types from 14 to 32 pins. The description below gives an overview of the complete range of peripherals proposed in this family.

These features make the ultra-low-power STM32L011x3/4 microcontrollers suitable for a wide range of applications:

- Gas/water meters and industrial sensors
- Healthcare and fitness equipment
- Remote control and user interface
- PC peripherals, gaming, GPS equipment
- Alarm system, wired and wireless sensors, video intercom

This STM32L011x3/4 datasheet should be read in conjunction with the STM32L0x1 reference manual (RM0377).

For information on the Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ core please refer to the Cortex<sup>®</sup>-M0+ Technical Reference Manual, available from the www.arm.com website.

*Figure 1* shows the general block diagram of the device family.



# 2 Description

The access line ultra-low-power STM32L011x3/4 family incorporates the high-performance Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ 32-bit RISC core operating at a 32 MHz frequency, high-speed embedded memories (up to 16 Kbytes of Flash program memory, 512 bytes of data EEPROM and 2 Kbytes of RAM) plus an extensive range of enhanced I/Os and peripherals.

The STM32L011x3/4 devices provide high power efficiency for a wide range of performance. It is achieved with a large choice of internal and external clock sources, an internal voltage adaptation and several low-power modes.

The STM32L011x3/4 devices offer several analog features, one 12-bit ADC with hardware oversampling, two ultra-low-power comparators, several timers, one low-power timer (LPTIM), three general-purpose 16-bit timers, one RTC and one SysTick which can be used as timebases. They also feature two watchdogs, one watchdog with independent clock and window capability and one window watchdog based on bus clock.

Moreover, the STM32L011x3/4 devices embed standard and advanced communication interfaces: one I2C, one SPI, one USART, and a low-power UART (LPUART).

The STM32L011x3/4 also include a real-time clock and a set of backup registers that remain powered in Standby mode.

The ultra-low-power STM32L011x3/4 devices operate from a 1.8 to 3.6 V power supply (down to 1.65 V at power down) with BOR and from a 1.65 to 3.6 V power supply without BOR option. They are available in the -40 to +125 °C temperature range. A comprehensive set of power-saving modes allows the design of low-power applications.







# 2.1 Device overview

Table 2. Ultra-low-power STM32L011x3/x4 device features and per	ipheral counts
---	----------------

Peripheral		STM32 L011D3	STM32 L011F3	STM32 L011E3	STM32 L011G3	STM32 L011K3	STM32 L011D4	STM32 L011F4	STM32 L011E4	STM32 L011G4	STM32 L011K4	
Flash (Kbyte	s)			8					16			
Data EEPROI	M (bytes)					51	2					
RAM (Kbytes	)					2						
Timers	General- purpose		2									
	LPTIM					1						
RTC/SYSTIC WWE						1/1/	1/1					
	SPI		1									
Communi-	l <sup>2</sup> C		1									
cation interfaces	USART	1										
	LPUART	1										
GPIOs		11	16	21	24	26/28 <sup>(1)</sup>	11	16	21	24	26/28 <sup>(1)</sup>	
Clocks: HSE <sup>(2)</sup> /LSE/HSI/MSI/LSI		1/1/1/1										
12b synchronized ADC Number of channels		1 4	1 7/9 <sup>(3)</sup>		1 10		1 4	1 7/9 <sup>(3)</sup>		1 10		
Comparators		2										
Max. CPU frequency		32 MHz										
Operating voltage		1.8 V to 3.6 V (down to 1.65 V at power-down) with BOR option 1.65 V to 3.6 V without BOR option										
Operating temperatures		Ambient temperature: -40 to +125 °C Junction temperature: -40 to +130 °C										
Packages		TSSOP 14	TSSOP/ UFQFPN 20	WLCSP 25	UFQFPN 28	LQFP/ UFQFPN 32	TSSOP 14	TSSOP/ UFQFPN 20	WLCSP 25	UFQFPN 28	LQFP/, UFQFPN 32	

1. The devices feature 26 and 28 GPIOs on LQFP32 and UFQFPN32, respectively.

2. HSE available only as external clock input (HSE bypass).

3. The devices feature 7 and 9 ADC channels on UFQFPN20 and TSSOP20, respectively.



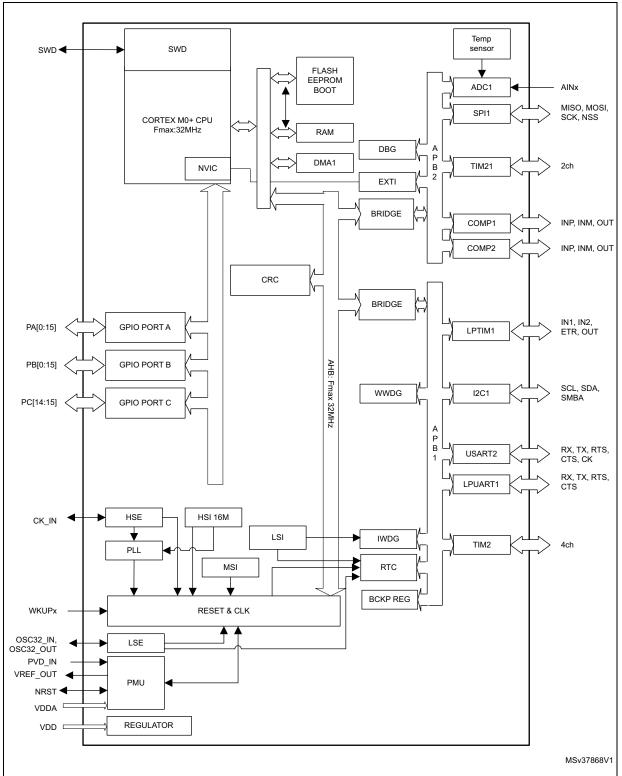


Figure 1. STM32L011x3/4 block diagram

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# 2.2 Ultra-low-power device continuum

The ultra-low-power family offers a large choice of core and features, from 8-bit proprietary core up to Arm<sup>®</sup> Cortex<sup>®</sup>-M4, including Arm<sup>®</sup> Cortex<sup>®</sup>-M3 and Arm<sup>®</sup> Cortex<sup>®</sup>-M0+. The STM32Lx series are the best choice to answer your needs in terms of ultra-low-power features. The STM32 Ultra-low-power series are the best solution for applications such as gas/water meter, keyboard/mouse or fitness and healthcare application. Several built-in features like LCD drivers, dual-bank memory, low-power Run mode, operational amplifiers, 128-bit AES, DAC, crystal-less USB and many other definitely help you building a highly cost optimized application by reducing BOM cost. STMicroelectronics, as a reliable and long-term manufacturer, ensures as much as possible pin-to-pin compatibility between all STM8Lx and STM32Lx on one hand, and between all STM32Lx and STM32Fx on the other hand. Thanks to this unprecedented scalability, your legacy application can be upgraded to respond to the latest market feature and efficiency requirements.



# **3** Functional overview

# 3.1 Low-power modes

The ultra-low-power STM32L011x3/4 supports dynamic voltage scaling to optimize its power consumption in Run mode. The voltage from the internal low-drop regulator that supplies the logic can be adjusted according to the system's maximum operating frequency and the external voltage supply.

There are three power consumption ranges:

- Range 1 (V<sub>DD</sub> range limited to 1.71-3.6 V), with the CPU running at up to 32 MHz
- Range 2 (full V<sub>DD</sub> range), with a maximum CPU frequency of 16 MHz
- Range 3 (full V<sub>DD</sub> range), with a maximum CPU frequency limited to 4.2 MHz

Seven low-power modes are provided to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

• Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs. Sleep mode power consumption at 16 MHz is about 1 mA with all peripherals off.

#### Low-power run mode

This mode is achieved with the multispeed internal (MSI) RC oscillator set to the lowspeed clock (max 131 kHz), execution from SRAM or Flash memory, and internal regulator in low-power mode to minimize the regulator's operating current. In Lowpower run mode, the clock frequency and the number of enabled peripherals are both limited.

#### • Low-power sleep mode

This mode is achieved by entering Sleep mode with the internal voltage regulator in low-power mode to minimize the regulator's operating current. In Low-power sleep mode, both the clock frequency and the number of enabled peripherals are limited; a typical example would be to have a timer running at 32 kHz.

When wakeup is triggered by an event or an interrupt, the system reverts to the Run mode with the regulator on.

#### • Stop mode with RTC

The Stop mode achieves the lowest power consumption while retaining the RAM and register contents and real time clock. All clocks in the  $V_{CORE}$  domain are stopped, the PLL, MSI RC, HSE and HSI RC oscillators are disabled. The LSE or LSI is still running. The voltage regulator is in the low-power mode.

Some peripherals featuring wakeup capability can enable the HSI RC during Stop mode to detect their wakeup condition.

The device can be woken up from Stop mode by any of the EXTI line, in  $3.5 \,\mu$ s, the processor can serve the interrupt or resume the code. The EXTI line source can be any GPIO. It can be the PVD output, the comparator 1 event or comparator 2 event



(if internal reference voltage is on), it can be the RTC alarm/tamper/timestamp/wakeup events, the USART/I2C/LPUART/LPTIM wakeup events.

#### • Stop mode without RTC

The Stop mode achieves the lowest power consumption while retaining the RAM and register contents. All clocks are stopped, the PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillator are disabled.

Some peripherals featuring wakeup capability can enable the HSI RC during Stop mode to detect their wakeup condition.

The voltage regulator is in the low-power mode. The device can be woken up from Stop mode by any of the EXTI line, in 3.5  $\mu$ s, the processor can serve the interrupt or resume the code. The EXTI line source can be any GPIO. It can be the PVD output, the comparator 1 event or comparator 2 event (if internal reference voltage is on). It can also be wakened by the USART/I2C/LPUART/LPTIM wakeup events.

#### • Standby mode with RTC

The Standby mode is used to achieve the lowest power consumption and real time clock. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSE bypass and HSI RC oscillators are also switched off. The LSE or LSI is still running. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32 KHz oscillator, RCC\_CSR register).

The device exits Standby mode in 60 µs when an external reset (NRST pin), an IWDG reset, a rising edge on one of the three WKUP pins, RTC alarm (Alarm A or Alarm B), RTC tamper event, RTC timestamp event or RTC Wakeup event occurs.

#### • Standby mode without RTC

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire  $V_{CORE}$  domain is powered off. The PLL, MSI RC, HSI and LSI RC, HSE and LSE crystal oscillator are also switched off. After entering Standby mode, the RAM and register contents are lost except for registers in the Standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE Crystal 32 KHz oscillator, RCC\_CSR register).

The device exits Standby mode in 60  $\mu$ s when an external reset (NRST pin) or a rising edge on one of the three WKUP pin occurs.

*Note:* The RTC, the IWDG, and the corresponding clock sources are not stopped automatically by entering Stop or Standby mode.



	Functionalities depending on the operating power supply range				
Operating power supply range	ADC operation	Dynamic voltage scaling range			
V <sub>DD</sub> = 1.65 to 1.71 V	ADC only, conversion time up to 570 ksps	Range 2 or range 3			
V <sub>DD</sub> = 1.71 to 1.8 V <sup>(1)</sup>	ADC only, conversion time up to 1.14 Msps	Range 1, range 2 or range 3			
$V_{DD}$ = 1.8 to 2.0 V <sup>(1)</sup>	Conversion time up to 1.14 Msps	Range1, range 2 or range 3			
V <sub>DD</sub> = 2.0 to 2.4 V	Conversion time up to 1.14 Msps	Range 1, range 2 or range 3			
V <sub>DD</sub> = 2.4 to 3.6 V	Conversion time up to 1.14 Msps	Range 1, range 2 or range 3			

	Table 3. Functionalities depending on the operating power supply range
--	--

 CPU frequency changes from initial to final must respect the condition: f<sub>CPU initial</sub> <4f<sub>CPU initial</sub>. It must also respect 5 µs delay between two changes. For example to switch from 4.2 MHz to 32 MHz, you can switch from 4.2 MHz to 16 MHz, wait 5 µs, then switch from 16 MHz to 32 MHz.

CPU frequency range	Dynamic voltage scaling range
16 MHz to 32 MHz (1ws) 32 kHz to 16 MHz (0ws)	Range 1
8 MHz to 16 MHz (1ws) 32 kHz to 8 MHz (0ws)	Range 2
32 kHz to 4.2 MHz (0ws)	Range 3



			Low-	Low-	w- Stop	Stop	5	Standby	
IPs	Run/Active	Sleep	power run	power sleep		Wakeup capability		Wakeup capability	
CPU	Y	-	Y	-	-	-	-	-	
Flash memory	0	0	0	0	-	-	-	-	
RAM	Y	Y	Y	Y	Y	-	-	-	
Backup registers	Y	Y	Y	Y	Y	-	Y	-	
EEPROM	0	0	0	0	-	-	-	-	
Brown-out reset (BOR)	0	0	0	0	0	0	0	0	
DMA	0	0	0	0	-	-	-	-	
Programmable Voltage Detector (PVD)	0	0	0	Ο	0	0	-	-	
Power-on/down reset (POR/PDR)	Y	Y	Y	Y	Y	Y	Y	Y	
High Speed Internal (HSI)	0	0	-	-	(3)	-	-	-	
High Speed External (HSE)	0	0	0	0	-	-	-	-	
Low Speed Internal (LSI)	0	0	0	0	0	-	0	-	
Low Speed External (LSE)	0	0	0	0	0	-	0	-	
Multi-Speed Internal (MSI)	0	0	Y	Y	-	-	-	-	
Inter-Connect Controller	Y	Y	Y	Y	Y	-	-	-	
RTC	0	0	0	0	0	0	0	-	
RTC Tamper	0	0	0	0	0	0	0	0	
Auto WakeUp (AWU)	0	0	0	0	0	-	0	0	
USART	0	0	0	0	O <sup>(4)</sup>	0	-	-	
LPUART	0	0	0	0	O <sup>(4)</sup>	0	-	-	
SPI	0	0	0	0	-		-	-	
I2C	0	0	0	0	O <sup>(5)</sup>	0	-	-	
ADC	0	0	-	-	-	-	-	-	
Temperature sensor	0	0	0	0	0	-	-	-	

Table 5. Functionalities depending on the working mode (from Run/active down to standby) <sup>(1)(2)</sup>



			Low-	Low-		Stop		Standby				
IPs	Run/Active	Sleep	power run	power sleep		Wakeup capability		Wakeup capability				
Comparators	0	0	0	0	0	0	-	-				
16-bit timers	0	0	0	0	-	-	-	-				
LPTIM	0	0	0	0	0	0	-	-				
IWDG	0	0	0	0	0	0	0	0				
WWDG	0	0	0	0	-	-	-	-				
SysTick Timer	0	0	0	0	-	-	-	-				
GPIOs	0	0	0	0	0	0	-	2 pins				
Wakeup time to Run mode	0 µs	6 CPU cycles	3 µs	7 CPU cycles	5 µs			65 µs				
					0.29 µA (No RTC) V <sub>DD</sub> =1.8 V			18 µA (No ) V <sub>DD</sub> =1.8 V				
Consumption	Down to	Down to	Down to			to Down to	0.54 µA (with RTC) V <sub>DD</sub> =1.8 V					-1 μΑ (with ) V <sub>DD</sub> =1.8 V
V <sub>DD</sub> =1.8 to 3.6 V (Typ)	128 µA/MHz (from Flash)	31 µA/MHz (from Flash)	7 μΑ	3.8 µA	0.67 µA (with							

### Table 5. Functionalities depending on the working mode (from Run/active down to standby) (continued)<sup>(1)(2)</sup>

1 Legend:

"Y" = Yes (enable). "O" = Optional, can be enabled/disabled by software) "-" = Not available

- 2. The consumption values given in this table are preliminary data given for indication. They are subject to slight changes.
- 3. Some peripherals with wakeup from Stop capability can request HSI to be enabled. In this case, HSI is woken up by the peripheral, and only feeds the peripheral which requested it. HSI is automatically put off when the peripheral does not need it anymore.
- 4. UART and LPUART reception is functional in Stop mode. It generates a wakeup interrupt on Start. To generate a wakeup on address match or received frame event, the LPUART can run on LSE clock while the UART has to wake up or keep running the HSI clock.
- 5. I2C address detection is functional in Stop mode. It generates a wakeup interrupt in case of address match. It will wake up the HSI during reception.

#### 3.2 Interconnect matrix

Several peripherals are directly interconnected. This allows autonomous communication between peripherals, thus saving CPU resources and power consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep, Low-power run, Low-power sleep and Stop modes.



Table 6. ST M32L011X3/4 peripherals interconnect matrix							
Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low- power run	Low- power sleep	Stop
COMPy	TIM2,TIM21	Timer input channel, trigger from analog signals comparison	Y	Y	Y	Y	-
COMPx Timer input cha LPTIM1 trigger from an		Timer input channel, trigger from analog signals comparison	Y	Y	Y	Y	Y
TIMx	TIMx	Timer triggered by other timer	Y	Y	Y	Y	-
RTC	TIM21	Timer triggered by Auto wake-up	Υ	Y	Y	Y	-
Ric	LPTIM1	Timer triggered by RTC event	Y	Y	Y	Y	Y
All clock source	TIMx	Clock source used as input channel for RC measurement and trimming	Y	Y	Y	Y	-
	TIMx	Timer input channel and trigger	Y	Y	Y	Y	-
GPIO	LPTIM1	Timer input channel and trigger	Υ	Y	Y	Y	Y
	ADC	Conversion trigger	Y	Y	Y	Y	-

 Table 6. STM32L011x3/4 peripherals interconnect matrix

# 3.3 Arm<sup>®</sup> Cortex<sup>®</sup>-M0+ core

The Cortex-M0+ processor is an entry-level 32-bit Arm Cortex processor designed for a broad range of embedded applications. It offers significant benefits to developers, including:

- a simple architecture that is easy to learn and program
- ultra-low power, energy-efficient operation
- excellent code density
- deterministic, high-performance interrupt handling
- upward compatibility with Cortex-M processor family
- platform security robustness.

The Cortex-M0+ processor is built on a highly area and power optimized 32-bit processor core, with a 2-stage pipeline von Neumann architecture. The processor delivers exceptional energy efficiency through a small but powerful instruction set and extensively optimized design, providing high-end processing hardware including a single-cycle multiplier.

The Cortex-M0+ processor provides the exceptional performance expected of a modern 32bit architecture, with a higher code density than other 8-bit and 16-bit microcontrollers.

Owing to its embedded Arm core, the STM32L011x3/4 are compatible with all Arm tools and software.



#### Nested vectored interrupt controller (NVIC)

The ultra-low-power STM32L011x3/4 embed a nested vectored interrupt controller able to handle up to 32 maskable interrupt channels and 4 priority levels.

The Cortex-M0+ processor closely integrates a configurable Nested Vectored Interrupt Controller (NVIC), to deliver industry-leading interrupt performance. The NVIC:

- includes a Non-Maskable Interrupt (NMI)
- provides zero jitter interrupt option
- provides four interrupt priority levels

The tight integration of the processor core and NVIC provides fast execution of Interrupt Service Routines (ISRs), dramatically reducing the interrupt latency. This is achieved through the hardware stacking of registers, and the ability to abandon and restart load-multiple and store-multiple operations. Interrupt handlers do not require any assembler wrapper code, removing any code overhead from the ISRs. Tail-chaining optimization also significantly reduces the overhead when switching from one ISR to another.

To optimize low-power designs, the NVIC integrates with the sleep modes, that include a deep sleep function that enables the entire device to enter rapidly stop or standby mode.

This hardware block provides flexible interrupt management features with minimal interrupt latency.

# 3.4 Reset and supply management

#### 3.4.1 Power supply schemes

- V<sub>DD</sub> = 1.65 to 3.6 V: external power supply for I/Os and the internal regulator. Provided externally through V<sub>DD</sub> pins.
- V<sub>SSA</sub>, V<sub>DDA</sub> = 1.65 to 3.6 V: external analog power supplies for ADC, reset blocks, RCs and PLL. V<sub>DDA</sub> and V<sub>SSA</sub> must be connected to V<sub>DD</sub> and V<sub>SS</sub>, respectively. On TSSOP14 package, V<sub>DDA</sub> is internally connected to V<sub>DD</sub>.

#### 3.4.2 Power supply supervisor

The devices feature an integrated ZEROPOWER power-on reset (POR)/power-down reset (PDR) that can be coupled with a brownout reset (BOR) circuitry.

Two versions are available:

- The version with BOR activated at power-on operates between 1.8 V and 3.6 V.
- The other version without BOR operates between 1.65 V and 3.6 V.

After the  $V_{DD}$  threshold is reached (1.65 V or 1.8 V depending on the BOR which is active or not at power-on), the option byte loading process starts, either to confirm or modify default thresholds, or to disable the BOR permanently: in this case, the VDD min value becomes 1.65 V (whatever the version, BOR active or not, at power-on).

When BOR is active at power-on, it ensures proper operation starting from 1.8 V whatever the power ramp-up phase before it reaches 1.8 V. When BOR is not active at power-up, the power ramp-up should guarantee that 1.65 V is reached on  $V_{DD}$  at least 1 ms after it exits the POR area.



Five BOR thresholds are available through option bytes, starting from 1.8 V to 3 V. To reduce the power consumption in Stop mode, it is possible to automatically switch off the internal reference voltage ( $V_{REFINT}$ ) in Stop mode. The device remains in reset mode when  $V_{DD}$  is below a specified threshold,  $V_{POR/PDR}$  or  $V_{BOR}$ , without the need for any external reset circuit.

Note: The start-up time at power-on is typically 3.3 ms when BOR is active at power-up, the startup time at power-on can be decreased down to 1 ms typically for devices with BOR inactive at power-up.

The devices feature an embedded programmable voltage detector (PVD) that monitors the  $V_{DD/VDDA}$  power supply and compares it to the  $V_{PVD}$  threshold. This PVD offers 7 different levels between 1.85 V and 3.05 V, chosen by software, with a step around 200 mV. An interrupt can be generated when  $V_{DD/VDDA}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD/VDDA}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

#### 3.4.3 Voltage regulator

The regulator has three operation modes: main (MR), low power (LPR) and power down.

- MR is used in Run mode (nominal regulation)
- LPR is used in the Low-power run, Low-power sleep and Stop modes
- Power down is used in Standby mode. The regulator output is high impedance, the kernel circuitry is powered down, inducing zero consumption but the contents of the registers and RAM are lost except for the standby circuitry (wakeup logic, IWDG, RTC, LSI, LSE crystal 32 KHz oscillator, RCC\_CSR).

#### 3.4.4 Boot modes

At startup, BOOT0 pin and nBOOT0, nBOOT1 and nBOOT\_SEL option bits are used to select one of three boot options:

- Boot from Flash memory
- Boot from System memory
- Boot from embedded RAM

The bootloader is located in System memory. It is used to reprogram the Flash memory by using SPI1 (PA4, PA7, PA13 and PA14 on TSSOP14 package or PA4, PA5, PA6 and PA7 on other packages) or USART2 (PA2, PA3 and PA9, PA10).

If the bootloader is activated (the bootloader is active on all empty devices due to the empty check mechanism), then the above mentioned bits are configured depending on whether SPI1 or USART2 functionality is used.

See STM32<sup>™</sup> microcontroller system memory boot mode AN2606 for details.



# 3.5 Clock management

The clock controller distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

#### Clock prescaler

To get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler.

#### • Safe clock switching

Clock sources can be changed safely on the fly in Run mode through a configuration register.

#### Clock management

To reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.

#### • System clock source

Three different clock sources can be used to drive the master clock SYSCLK:

- 0-32 MHz high-speed external (HSE bypass), that can supply a PLL
- 16 MHz high-speed internal RC oscillator (HSI), trimmable by software, that can supply a PLL
- Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 7 frequencies (65 kHz, 131 kHz, 262 kHz, 524 kHz, 1.05 MHz, 2.1 MHz, 4.2 MHz). When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be trimmed by software down to a ±0.5% accuracy.

#### Auxiliary clock source

Two ultra-low-power clock sources that can be used to drive the real-time clock:

- 32.768 kHz low-speed external crystal (LSE)
- 37 kHz low-speed internal RC (LSI), also used to drive the independent watchdog. The LSI clock can be measured using the high-speed internal RC oscillator for greater precision.

#### • RTC clock sources

The LSI, LSE or HSE sources can be chosen to clock the RTC, whatever the system clock.

#### • Startup clock

After reset, the microcontroller restarts by default with an internal 2 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.

#### • Clock security system (CSS)

This feature can be enabled by software. If an LSE clock failure occurs, it provides an interrupt or wakeup event which is generated assuming it has been previously enabled. This feature is not available on the HSE clock.

#### • Clock-out capability (MCO: microcontroller clock output)

It outputs one of the internal clocks for external use by the application.

Several prescalers allow the configuration of the AHB frequency, each APB (APB1 and APB2) domains. The maximum frequency of the AHB and the APB domains is 32 MHz. See *Figure 2* for details on the clock tree.



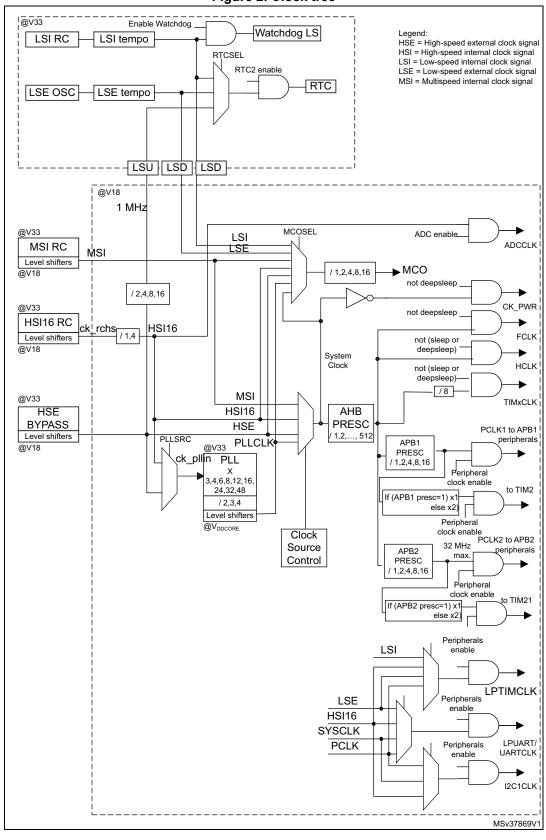


Figure 2. Clock tree



# 3.6 Low-power real-time clock and backup registers

The real time clock (RTC) and the 5 backup registers are supplied in all modes including standby mode. The backup registers are five 32-bit registers used to store 20 bytes of user application data. They are not reset by a system reset, or when the device wakes up from Standby mode.

The RTC is an independent BCD timer/counter. Its main features are the following:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format
- Automatically correction for 28, 29 (leap year), 30, and 31 day of the month
- Two programmable alarms with wake up from Stop and Standby mode capability
- Periodic wakeup from Stop and Standby with programmable resolution and period
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 1 ppm resolution, to compensate for quartz crystal inaccuracy
- 2 anti-tamper detection pins with programmable filter. The MCU can be woken up from Stop and Standby modes on tamper event detection.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event. The MCU can be woken up from Stop and Standby modes on timestamp event detection.

The RTC clock sources can be:

- A 32.768 kHz external crystal
- A resonator or oscillator
- The internal low-power RC oscillator (typical frequency of 37 kHz)
- The high-speed external clock

# 3.7 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions, and can be individually remapped using dedicated alternate function registers. All GPIOs are high current capable. Each GPIO output, speed can be slowed (40 MHz, 10 MHz, 2 MHz, 400 kHz). The alternate function configuration of I/Os can be locked if needed following a specific sequence in order to avoid spurious writing to the I/O registers. The I/O controller is connected to a dedicated IO bus with a toggling speed of up to 32 MHz.

The BOOT0 pin is shared with PB9 GPIO pin. This pin is an input-only pin. If nBOOT\_SEL option bit is reset, sampling this pin on NRST rising edge gives the internal BOOT0 state. This pin then works as PB9 pin. The input voltage characteristics of this pin are specific for BOOT0 pin type (see *Table 50: I/O static characteristics*).

# Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 26 edge detector lines used to generate interrupt/event requests. Each line can be individually configured to select the trigger event



(rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 38 GPIOs can be connected to the 16 configurable interrupt/event lines. The 10 other lines are connected to PVD, RTC, USART, I2C, LPUART, LPTIM or comparator events.

### 3.8 Memories

The STM32L011x3/4 devices have the following features:

- 2 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states. With the enhanced bus matrix, operating the RAM does not lead to any performance penalty during accesses to the system bus (AHB and APB buses).
- The non-volatile memory is divided into three arrays:
  - 8 or 16 Kbytes of embedded Flash program memory
  - 512 bytes of data EEPROM
  - Information block containing 32 user and factory options bytes plus 4 Kbytes of system memory

The user options bytes are used to write-protect or read-out protect the memory (with 4 Kbyte granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no protection
- Level 1: memory readout protected.

The Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected

• Level 2: chip readout protected, debug features (Cortex-M0+ serial wire) and boot in RAM selection disabled (debugline fuse)

The whole non-volatile memory embeds the error correction code (ECC) feature.

# 3.9 Direct memory access (DMA)

The flexible 5-channel, general-purpose DMA is able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA controller supports circular buffer management, avoiding the generation of interrupts when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with software trigger support for each channel. Configuration is done by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI, I<sup>2</sup>C, USART, LPUART, general-purpose timers, and ADC.



# 3.10 Analog-to-digital converter (ADC)

A native 12-bit, extended to 16-bit through hardware oversampling, analog-to-digital converter is embedded into STM32L011x3/4 devices. It has up to 10 external channels and 2 internal channels (temperature sensor, voltage reference). Three channels, PA0, PA4 and PA5, are fast channels, while the others are standard channels.

The ADC performs conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC frequency is independent from the CPU frequency, allowing maximum sampling rate of 1.14 MSPS even with a low CPU speed. The ADC consumption is low at all frequencies (~25  $\mu$ A at 10 kSPS, ~200  $\mu$ A at 1MSPS). An auto-shutdown function guarantees that the ADC is powered off except during the active conversion phase.

The ADC can be served by the DMA controller. It can operate from a supply voltage down to 1.65 V.

The ADC features a hardware oversampler up to 256 samples, this improves the resolution to 16 bits (see AN2668).

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions and timers.

# 3.11 Temperature sensor

The temperature sensor ( $T_{SENSE}$ ) generates a voltage  $V_{SENSE}$  that varies linearly with temperature.

The temperature sensor is internally connected to the ADC\_IN18 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode (see *Table 57: Temperature sensor calibration values*).

### 3.11.1 Internal voltage reference (V<sub>REFINT</sub>)

The internal voltage reference ( $V_{REFINT}$ ) provides a stable (bandgap) voltage output for the ADC and Comparators.  $V_{REFINT}$  is internally connected to the ADC\_IN17 input channel. It enables accurate monitoring of the  $V_{DD}$  value (since no external voltage,  $V_{REF+}$ , is available for ADC). The precise voltage of  $V_{REFINT}$  is individually measured for each part by ST during production test and stored in the system memory area (see *Table 20: Embedded internal reference voltage calibration values*). It is accessible in read-only mode.



### 3.12 Ultra-low-power comparators and reference voltage

The STM32L011x3/4 embed two comparators sharing the same current bias and reference voltage. The reference voltage can be internal or external (coming from an I/O).

- One comparator with ultra low consumption
- One comparator with rail-to-rail inputs, fast or slow mode.
- The threshold can be one of the following:
  - External I/O pins
  - Internal reference voltage (V<sub>REFINT</sub>)
  - submultiple of Internal reference voltage(1/4, 1/2, 3/4) for the rail to rail comparator.

Both comparators can wake up the devices from Stop mode, and be combined into a window comparator.

The internal reference voltage is available externally via a low-power / low-current output buffer (driving current capability of 1  $\mu$ A typical).

# 3.13 System configuration controller

The system configuration controller provides the capability to remap some alternate functions on different I/O ports.

The highly flexible routing interface allows the application firmware to control the routing of different I/Os to the TIM2, TIM21 and LPTIM1 timer input captures. It also controls the routing of internal analog signals to the ADC, COMP1 and COMP2 and the internal reference voltage  $V_{\text{REFINT}}$ .

# 3.14 Timers and watchdogs

The ultra-low-power STM32L011x3/4 devices include two general-purpose timers, one low-power timer (LPTIM1), two watchdog timers and the SysTick timer.

*Table 7* compares the features of the general-purpose and basic timers.

Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
TIM2	16-bit	Up, down, up/down	Any integer between 1 and 65536	Yes	4	No
TIM21	16-bit	Up, down, up/down	Any integer between 1 and 65536	No	2	No

Table 7. Timer feature comparison



# 3.14.1 General-purpose timers (TIM2, TIM21)

There are three synchronizable general-purpose timers embedded in the STM32L011x3/4 devices (see *Table 7* for differences).

#### TIM2

TIM2 is based on 16-bit auto-reload up/down counter. It includes a 16-bit prescaler. It features four independent channels each for input capture/output compare, PWM or one-pulse mode output.

The TIM2 general-purpose timers can work together or with the TIM21 general-purpose timer via the Timer Link feature for synchronization or event chaining. Its counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs.

TIM2 has independent DMA request generation.

This timer is capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

#### TIM21

TIM21 is based on a 16-bit auto-reload up/down counter. It includes a 16-bit prescaler. It has two independent channels for input capture/output compare, PWM or one-pulse mode output. It can work together and be synchronized with TIM2 full-featured general-purpose timer.

It can also be used as simple timebase and be clocked by the LSE clock source (32.768 kHz) to provide independent timebase from the main CPU clock.

### 3.14.2 Low-power Timer (LPTIM)

The low-power timer has an independent clock and is running also in Stop mode if it is clocked by LSE, LSI or an external clock. It is able to wakeup the devices from Stop mode.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous / one shot mode
- Selectable software / hardware input trigger
- Selectable clock source
  - Internal clock source: LSE, LSI, HSI or APB clock
  - External clock source over LPTIM1 input (working even with no internal clock source running, used by the Pulse Counter Application)
  - Programmable digital glitch filter
- Encoder mode

#### 3.14.3 SysTick timer

This timer is dedicated to the OS, but could also be used as a standard downcounter. It is based on a 24-bit downcounter with autoreload capability and a programmable clock source. It features a maskable system interrupt generation when the counter reaches '0'.

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### 3.14.4 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 37 kHz internal RC and, as it operates independently of the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes. The counter can be frozen in debug mode.

### 3.14.5 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

# 3.15 Communication interfaces

# 3.15.1 I<sup>2</sup>C bus

One I<sup>2</sup>C interface (I2C1) can operate in multimaster or slave modes. The I<sup>2</sup>C interface can support Standard mode (Sm, up to 100 kbit/s), Fast mode (Fm, up to 400 kbit/s) and Fast Mode Plus (Fm+, up to 1 Mbit/s) with 20 mA output drive on some I/Os.

The I<sup>2</sup>C interface supports 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (2 addresses, 1 with configurable mask). They also include programmable analog and digital noise filters.

	Analog filter	Digital filter
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2C peripheral clocks
Benefits	Available in Stop mode	<ol> <li>Extra filtering capability vs. standard requirements.</li> <li>Stable length</li> </ol>
Drawbacks	Variations depending on temperature, voltage, process	Wakeup from Stop on address match is not available when digital filter is enabled.

In addition, I2C1 provides hardware support for SMBus 2.0 and PMBus 1.1: ARP capability, Host notify protocol, hardware CRC (PEC) generation/verification, timeouts verifications and ALERT protocol management. I2C1 also has a clock domain independent from the CPU clock, allowing the I2C1 to wake up the MCU from Stop mode on address match.

The I2C interface can be served by the DMA controller.

Refer to *Table 9* for the supported modes and features of I2C interface.



I2C features <sup>(1)</sup>	I2C1
7-bit addressing mode	Х
10-bit addressing mode	Х
Standard mode (up to 100 kbit/s)	Х
Fast mode (up to 400 kbit/s)	Х
Fast Mode Plus with 20 mA output drive I/Os (up to 1 Mbit/s)	X <sup>(2)</sup>
Independent clock	Х
SMBus	Х
Wakeup from STOP	Х

### Table 9. STM32L011x3/4 I<sup>2</sup>C implementation

1. X = supported.

2. See Table 13: Pin definitions on page 37 for the list of I/Os that feature Fast Mode Plus capability

### 3.15.2 Universal synchronous/asynchronous receiver transmitter (USART)

The USART interface (USART2) is able to communicate at speeds of up to 4 Mbit/s.

It provides hardware management of the CTS, RTS and RS485 driver enable (DE) signals, multiprocessor communication mode and single-wire half-duplex communication mode. USART2 also supports Smartcard communication (ISO 7816, T=0 protocol) and IrDA SIR ENDEC.

USART2 interface can be served by the DMA controller.

*Table 10* for the supported modes and features of USART interface.

USART modes/features <sup>(1)</sup>	USART2
Hardware flow control for modem	Х
Continuous communication using DMA	Х
Multiprocessor communication	Х
Synchronous mode	-
Smartcard mode	Х
Single-wire half-duplex communication	Х
IrDA SIR ENDEC block	Х
LIN mode	-
Dual clock domain and wakeup from Stop mode	-
Receiver timeout interrupt	-
Modbus communication	-
Auto baud rate detection (4 modes)	-
Driver Enable	Х

#### Table 10. USART implementation

1. X = supported.



### 3.15.3 Low-power universal asynchronous receiver transmitter (LPUART)

The devices embed one Low-power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock, and can wake up the system from Stop mode, using baudrates up to 46 Kbaud. The Wakeup events from Stop mode are programmable and can be:

- Start bit detection
- Or any received data frame
- Or a specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

LPUART interface can be served by the DMA controller.

### 3.15.4 Serial peripheral interface (SPI)

The SPI is able to communicate at up to 16 Mbits/s in slave and master modes in full-duplex and half-duplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes.

The SPI can be served by the DMA controller.

Refer to *Table 11* for the supported modes and features of SPI interface.

SPI features <sup>(1)</sup>	SPI1
Hardware CRC calculation	Х
I2S mode	-
TI mode	Х

#### Table 11. SPI implementation

1. X = supported.

# 3.16 Cyclic redundancy check (CRC) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at linktime and stored at a given memory location.

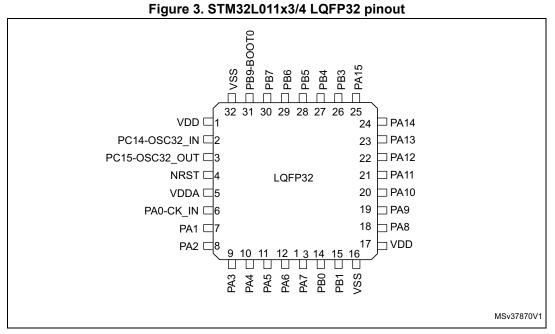


# 3.17 Serial wire debug port (SW-DP)

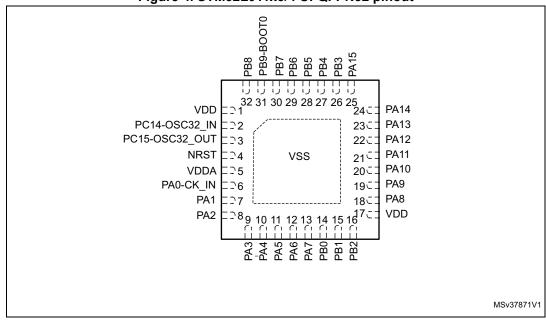
An Arm SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.



# 4 Pin descriptions

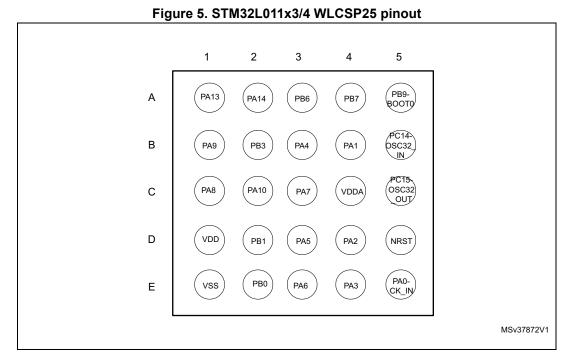


1. The above figure shows the package top view.



#### Figure 4. STM32L011x3/4 UFQFPN32 pinout





1. The above figure shows the package top view.

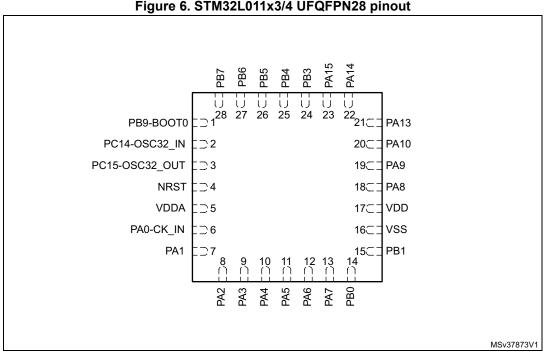
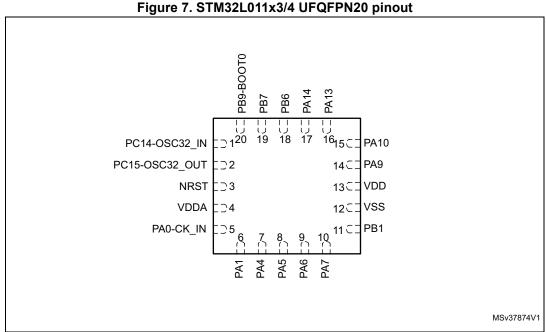
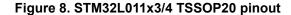


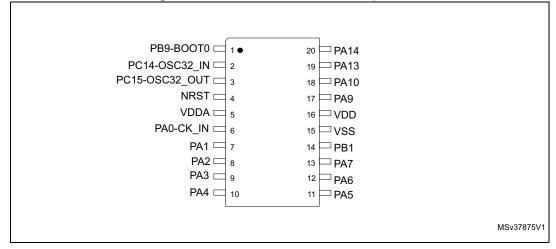
Figure 6. STM32L011x3/4 UFQFPN28 pinout





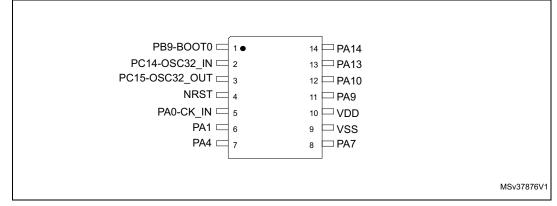
The above figure shows the package top view.











Nar	ne	Abbreviation	Definition
Pin name		Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type		S	Supply pin
		I	Input only pin
		I/O	Input / output pin
I/O structure		FT	5 V tolerant I/O
		FTf	5 V tolerant I/O, FM+ capable
		TTa	3.3 V tolerant I/O directly connected to the ADC
		TC	Standard 3.3V I/O
		В	Dedicated BOOT0 pin
		RST	Bidirectional reset pin with embedded weak pull-up resistor
Notes		Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset.	
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers	
	Additional functions	Functions directly selected/enabled through peripheral registers	



		Pin	num	ber							Pin functions				
TSSOP14	UFQFPN20	TSSOP20	UFQFPN28	LQFP32	UFQFPN32 <sup>(1)</sup>	WLCSP25	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions			
2	1	2	2	2	2	В5	PC14- OSC32_IN	I/O	FT	-	-	OSC32_IN			
3	2	3	3	3	3	C5	PC15- OSC32_OUT	I/O	тс	-	-	OSC32_OUT			
4	3	4	4	4	4	D5	NRST	I/O	RST	(2)	-	-			
10	4	5	5	5	5	C4	VDDA	S	-	(3)(4)	-	-			
5	5	6	6	6	6	E5	PA0-CK_IN	I/O	TTa	-	USART2_RX, LPTIM1_IN1, TIM2_CH1, USART2_CTS, TIM2_ETR, LPUART1_RX, COMP1_OUT	COMP1_INM, ADC_IN0, RTC_TAMP2/WKU P1/CK_IN			
6	6	7	7	7	7	B4	PA1	I/O	FT	-	EVENTOUT, LPTIM1_IN2, TIM2_CH2, I2C1_SMBA, USART2_RTS_DE, TIM21_ETR, LPUART1_TX	COMP1_INP, ADC_IN1			
-	-	8	8	8	8	D4	PA2	I/O	ТТа	-	TIM21_CH1, TIM2_CH3, USART2_TX, LPUART1_TX, COMP2_OUT	COMP2_INM, ADC_IN2, RTC_TAMP3/RTC_ TS/RTC_OUT/WKU P3			
-	-	9	9	9	9	E4	PA3	I/O	FT	-	TIM21_CH2, TIM2_CH4, USART2_RX, LPUART1_RX	COMP2_INP, ADC_IN3			
7	7	10	10	10	10	В3	PA4	I/O	ТТа	-	SPI1_NSS, LPTIM1_IN1, LPTIM1_ETR, I2C1_SCL, USART2_CK, TIM2_ETR, LPUART1_TX, COMP2_OUT	COMP1_INM, COMP2_INM, ADC_IN4			



Table 13. Pin definitions (c	continued)
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		Pin	num	ber							Pin fur	octions
TSSOP14	UFQFPN20	TSSOP20	UFQFPN28	LQFP32	UFQFPN32 <sup>(1)</sup>	WLCSP25	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	8	11	11	11	11	D3	PA5	I/O	ТТа	-	SPI1_SCK, LPTIM1_IN2, TIM2_ETR, TIM2_CH1	COMP1_INM, COMP2_INM, ADC_IN5
-	9	12	12	12	12	E3	PA6	PA6 I/O FT		-	SPI1_MISO, LPTIM1_ETR, LPUART1_CTS, EVENTOUT, COMP1_OUT	ADC_IN6
8	10	13	13	13	13	С3	PA7	I/O	FT	-	SPI1_MOSI, LPTIM1_OUT, USART2_CTS, TIM21_ETR, EVENTOUT, COMP2_OUT	COMP2_INP, ADC_IN7
-	-	-	14	14	14	E2	PB0	I/O	FT	-	EVENTOUT, SPI1_MISO, TIM2_CH2, USART2_RTS_DE, TIM2_CH3	ADC_IN8, VREF_OUT
-	11	14	15	15	15	D2	PB1	I/O	FT	-	USART2_CK, SPI1_MOSI, LPTIM1_IN1, LPUART1_RTS_DE , TIM2_CH4	ADC_IN9, VREF_OUT
-	-	-	-	-	16	-	PB2	I/O	FT	-	LPTIM1_OUT	-
9	12	15	16	16	-	E1	VSS	S	-	(5)	-	-
10	13	16	17	17	17	D1	VDD	S	-	(6)	-	-
-	-	-	18	18	18	C1	PA8	I/O	FT	-	MCO, LPTIM1_IN1, EVENTOUT, USART2_CK, TIM2_CH1	-
11	14	17	19	19	19	B1	PA9	I/O	FTf	-	MCO, I2C1_SCL, LPTIM1_OUT, USART2_TX, TIM21_CH2, COMP1_OUT	-



		Pin	num	ber						, Pin fur	ictions	
TSSOP14	UFQFPN20	TSSOP20	UFQFPN28	LQFP32	UFQFPN32 <sup>(1)</sup>	WLCSP25	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
12	15	18	20	20	20	C2	PA10	I/O	FTf	(7)	TIM21_CH1, I2C1_SDA, RTC_REFIN, USART2_RX, TIM2_CH3, COMP1_OUT	-
-	-	-	-	21	21	-	PA11	I/O	FT	-	SPI1_MISO, LPTIM1_OUT, EVENTOUT, USART2_CTS, TIM21_CH2, COMP1_OUT	-
-	-	-	-	22	22	-	PA12	I/O	FT	-	SPI1_MOSI, EVENTOUT, USART2_RTS_DE, COMP2_OUT	-
13	16	19	21	23	23	A1	PA13	I/O	FTf	-	SWDIO, LPTIM1_ETR, I2C1_SDA, SPI1_SCK, LPUART1_RX, COMP1_OUT	-
14	17	20	22	24	24	A2	PA14	I/O	FT	(7)	SWCLK, LPTIM1_OUT, I2C1_SMBA, USART2_TX, SPI1_MISO, LPUART1_TX, COMP2_OUT	-
-	-	-	23	25	25	-	PA15	I/O	FT	-	SPI1_NSS, TIM2_ETR, EVENTOUT, USART2_RX, TIM2_CH1	-
-	-	-	24	26	26	B2	PB3	I/O	FT	-	SPI1_SCK, TIM2_CH2, EVENTOUT	COMP2_INM
-	-	-	25	27	27	-	PB4	I/O	FT	-	SPI1_MISO, EVENTOUT	COMP2_INP



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Table 13. Pin definitions (c	continued)
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		Pin	num	ber	-						Pin fur	nctions
TSSOP14	UFQFPN20	TSSOP20	UFQFPN28	LQFP32	UFQFPN32 <sup>(1)</sup>	WLCSP25	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
-	-	-	26	28	28	-	PB5	I/O	FT	-	SPI1_MOSI, LPTIM1_IN1, I2C1_SMBA, TIM21_CH1	COMP2_INP
-	18	-	27	29	29	A3	PB6	I/O	FTf	-	USART2_TX, I2C1_SCL, LPTIM1_ETR, TIM2_CH3, LPUART1_TX	COMP2_INP
-	19	-	28	30	30	A4	PB7	I/O	FTf	-	USART2_RX, I2C1_SDA, LPTIM1_IN2, TIM2_CH4, LPUART1_RX	COMP2_INP, VREF_PVD_IN
1	20	1	1	31	31	A5	PB9-BOOT0	I	В	-	-	BOOT0 (Boot memory selection)
-	-	-	-	-	32	-	PB8	I/O	FTf	-	USART2_TX, EVENTOUT, I2C1_SCL, SPI1_NSS	-
-	-	-	-	32	-	-	VSS	S	-	(5)	-	-
-	-	-	-	1	1	-	VDD	S	-	(6)	-	-

1. V<sub>SS</sub> pins are connected to the exposed pad (see *Figure 35: UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package outline*).

2. Device reset input/internal reset output (active low).

- 3. Analog power supply.
- 4. On TSSOP14 package,  $V_{\text{DDA}}$  is internally connected to  $V_{\text{DD}}.$
- 5. Digital and analog ground.
- 6. Digital power supply.
- 7. PA14 pin on TSSOP14 package acts as an output pin when the embedded bootloader is active (SPI1\_MISO). On empty devices (devices from factory), the bootloader is active due to the empty check mechanism (refer to RM0377 reference manual). PA14 pin also acts as SWCLK. When programming devices in TSSOP14 for the first time, it is necessary to use the "connect under reset" method and the SWD interface to disable the bootloader by driving this PA14/SWCLK pin.



	AF7	COMP1/2	COMP1_OUT	ı	COMP2_OUT	ı	COMP2_OUT	I	COMP1_OUT	COMP2_OUT	I	COMP1_OUT	COMP1_OUT	COMP1_OUT	COMP2_OUT	COMP1_OUT	COMP2_OUT	
	AF6	LPUART1/EVE VENTOUT	LPUART1_RX	LPUART1_TX	LPUART1_TX	LPUART1_RX	LPUART1_TX	I	EVENTOUT	EVENTOUT	I	I	I	I		LPUART1_RX	LPUART1_TX	
	AF5	SPI1/TIM2/21	TIM2_ETR	TIM21_ETR	I	I	TIM2_ETR	TIM2_CH1	I	TIM21_ETR	TIM2_CH1	TIM21_CH2	TIM2_CH3	TIM21_CH2	I	SPI1_SCK	SPI1_MISO	TIM2_CH1
nctions	AF4	I2C1/USART2/L PUART1/ EVENTOUT	USART2_CTS	USART2_RTS_ DE	USART2_TX	USART2_RX	USART2_CK	I	LPUART1_CTS	USART2_CTS	USART2_CK	USART2_TX	USART2_RX	USART2_CTS	USART2_RTS_ DE	I	USART2_TX	USART2_RX
Table 14. Alternate functions	AF3	I2C1/LPTIM/ EVENTOUT	1	I2C1_SMBA	ı	1	I2C1_SCL	1	ı	ı	EVENTOUT	I	ı	1	-	I2C1_SDA	I2C1_SMBA	EVENTOUT
Table	AF2	LPUART1/ LPTIM/TIM2/ EVENTOUT/ SYS_AF	TIM2_CH1	TIM2_CH2	TIM2_CH3	TIM2_CH4	LPTIM1_ETR	TIM2_ETR	ı	ı	LPTIM1_IN1	LPTIM1_OUT	RTC_REFIN	EVENTOUT	EVENTOUT	ı	I	TIM2_ETR
	AF1	SP11/I2C1/ LPTIM	LPTIM1_IN1	LPTIM1_IN2	I	I	LPTIM1_IN1	LPTIM1_IN2	LPTIM1_ETR	LPTIM1_OUT	I	I2C1_SCL	I2C1_SDA	LPTIM1_OUT	T	LPTIM1_ETR	LPTIM1_OUT	
	AF0	SPI1/USART2/ TIM21/ EVENTOUT/ SYS_AF	USART2_RX	EVENTOUT	TIM21_CH1	TIM21_CH2	SPI1_NSS	SPI1_SCK	SPI1_MISO	SPI1_MOSI	MCO	MCO	TIM21_CH1	SPI1_MISO	SPI1_MOSI	OIDMS	SWCLK	SPI1_NSS
	Ports		PA0	PA1	PA2	PA3	PA4	PA5	PA6	PA7	PA8	PA9	PA10	PA11	PA12	PA13	PA14	PA15
		ď								< +100								

Pin descriptions

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AF3     AF4     AF5     AF6     AF7       I2C1/LPTIM/ EVENTOUT     I2C1/USART2/L PUART1/ EVENTOUT     I2C1/USART2/L SPI1/TIM2/21     LPUART1/EVE VENTOUT     COMP1/2	- USART2_RTS_ TIM2_CH3	- LPUART1_RTS_ TIM2_CH4		- EVENTOUT		I2C1_SMBA - TIM21_CH1	- TIM2_CH3 LPUART1_TX -	- TIM2_CH4 LPUART1_RX -	- I2C1_SCL SPI1_NSS		•	
-			LPTIM1_OUT	TIM2_CH2 - EVENTOUT	EVENTOUT	LPTIM1_IN1 I2C1_SMBA -	LPTIM1_ETR	LPTIM1_IN2	EVENTOUT - I2C1_SCL			
SPI1_MISO SPI1_MOSI			ſ	I	·		I2C1_SCL	I2C1_SDA	I	I	I	•
EVENTOUT USART2_CK	USART2_CK	ı		SPI1_SCK	SPI1_MISO	SPI1_MOSI	USART2_TX	USART2_RX	USART2_TX	•	ı	'
PB0 PB1	PB1		PB2	PB3	Port B PB4	PB5	PB6	PB7	PB8	PB9	PC14	PC15

# Table 14. Alternate functions (continued)

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# 5 Memory mapping

Refer to the product line reference manual for details on the memory mapping as well as the boundary addresses for all peripherals.



# 6 Electrical characteristics

### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

### 6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_Amax$  (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\sigma$ ).

### 6.1.2 Typical values

Unless otherwise specified, typical data are based on T<sub>A</sub> = 25 °C, V<sub>DD</sub> = 3.6 V (for the 1.65 V  $\leq$  V<sub>DD</sub>  $\leq$  3.6 V voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\sigma$ ).

### 6.1.3 Typical curves

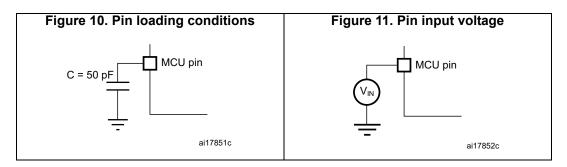
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in *Figure 10*.

### 6.1.5 Pin input voltage

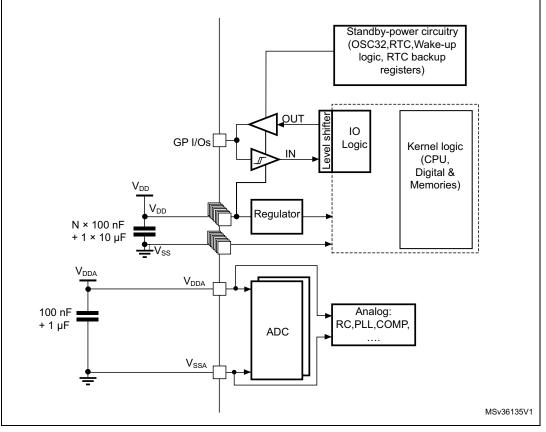
The input voltage measurement on a pin of the device is described in Figure 11.



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# 6.1.6 Power supply scheme



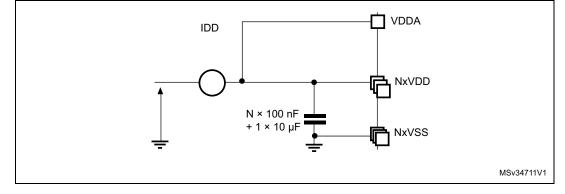
### Figure 12. Power supply scheme

1. On TSSOP14 package,  $V_{DDA}$  is internally connected to  $V_{DD}$ .

2.  $V_{SSA}$  is internally connected to  $V_{SS}$  on all packages.

### 6.1.7 Current consumption measurement

### Figure 13. Current consumption measurement scheme





### 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in Table 15: Voltage characteristics, Table 16: Current characteristics, and Table 17: Thermal characteristics may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 Qualification Standard. Extended mission profiles are available on demand.

Symbol	Ratings	Min	Мах	Unit
V <sub>DD</sub> -V <sub>SS</sub>	External main supply voltage (including V <sub>DDA</sub> , V <sub>DD</sub> ) <sup>(1)</sup>	-0.3	4.0	
	Input voltage on FT and FTf pins	V <sub>SS</sub> –0.3	V <sub>DD</sub> +4.0	
V <sub>IN</sub> <sup>(2)</sup>	Input voltage on TC pins	V <sub>SS</sub> –0.3	4.0	V
VIN Y	Input voltage on BOOT0	V <sub>SS</sub>	V <sub>DD</sub> +4.0	
	Input voltage on any other pin	V <sub>SS</sub> -0.3	4.0	
$ \Delta V_{DD} $	Variations between different $V_{DDx}$ power pins	-	50	
V <sub>DDA</sub> -V <sub>DDx</sub>	Variations between any $V_{DDx}$ and $V_{DDA}$ power $pins^{(3)}$	-	300	mV
$ \Delta V_{SS} $	Variations between all different ground pins	-	50	
V <sub>ESD(HBM)</sub> Electrostatic discharge voltage (human body model)		see Secti	ion 6.3.11	

Table 15	Voltage	characteristics
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All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

V<sub>IN</sub> maximum must always be respected. Refer to Table 16 for maximum allowed injected current values. 2.

It is recommended to power  $V_{DD}$  and  $V_{DDA}$  from the same source. A maximum difference of 300 mV between  $V_{DD}$  and  $V_{DDA}$  can be tolerated during power-up and device operation. its value does not need to respect this rule. 3.



Symbol	Ratings	Max.	Unit
$\Sigma I_{VDD}^{(2)}$	Total current into sum of all $V_{DD}$ power lines (source) <sup>(1)</sup>	105	
$\Sigma I_{VSS}^{(2)}$	Total current out of sum of all $V_{SS}$ ground lines (sink) <sup>(1)</sup>	105	
I <sub>VDD(PIN)</sub>	Maximum current into each V <sub>DD</sub> power pin (source) <sup>(1)</sup>	100	
I <sub>VSS(PIN)</sub>	Maximum current out of each $V_{SS}$ ground pin $(sink)^{(1)}$	100	
	Output current sunk by any I/O and control pin except FTf pins	16	
I <sub>IO</sub>	Output current sunk by FTf pins	22	
	Output current sourced by any I/O and control pin	-16	
<b>SI</b> (3)	Total output current sunk by sum of all IOs and control $\ensuremath{pins^{(4)}}$	45	mA
ΣΙ <sub>ΙΟ(ΡΙΝ)</sub> <sup>(3)</sup>	Total output current sourced by sum of all IOs and control pins	-45	
21	Total output current sunk by sum of all IOs and control $\ensuremath{pins}^{(2)}$	90	
ΣΙ <sub>ΙΟ(ΡΙΝ)</sub>	Total output current sourced by sum of all IOs and control $pins^{(2)}$	-90	
	Injected current on FT, FFf, RST and B pins	-5/+0 <sup>(5)</sup>	
I <sub>INJ(PIN)</sub>	Injected current on TC pin	± 5 <sup>(6)</sup>	
ΣΙ <sub>INJ(PIN)</sub>	Total injected current (sum of all I/O and control pins) <sup>(7)</sup>	± 25	

### Table 16. Current characteristics

1. All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

 This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.

- 3. These values apply only to STM32L011GxUx part number (UFQFPN28 package).
- 4. This current consumption must be correctly distributed over all I/Os and control pins. In particular, it must be located the closest possible to the couple of supply and ground, and distributed on both sides.
- Positive current injection is not possible on these I/Os. A negative injection is induced by V<sub>IN</sub><V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to Table 15 for maximum allowed input voltage values.
- A positive injection is induced by V<sub>IN</sub> > V<sub>DD</sub> while a negative injection is induced by V<sub>IN</sub> < V<sub>SS</sub>. I<sub>INJ(PIN)</sub> must never be exceeded. Refer to *Table 15: Voltage characteristics* for the maximum allowed input voltage values.
- 7. When several inputs are submitted to a current injection, the maximum  $\Sigma I_{INJ(PIN)}$  is the absolute sum of the positive and negative injected currents (instantaneous values).

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	–65 to +150	°C
TJ	Maximum junction temperature	150	°C

### Table 17. Thermal characteristics



# 6.3 Operating conditions

## 6.3.1 General operating conditions

Table 18. General operating conditions	Table 18	. General	operating	conditions
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Symbol	Parameter	Conditions	Min	Max	Unit	
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	32		
f <sub>PCLK1</sub>	Internal APB1 clock frequency	-	0	32	MHz	
f <sub>PCLK2</sub>	Internal APB2 clock frequency	-	0	32		
		BOR detector disabled	1.65	3.6		
V <sub>DD</sub>	Standard operating voltage	BOR detector enabled, at power on	1.8	3.6	V	
		BOR detector disabled, after power on	1.65	3.6		
V <sub>DDA</sub>	Analog operating voltage (all features)	3.6	V			
	Input voltage on FT, FTf and RST pins <sup>(2)</sup>	$2.0 \text{ V} \le \text{V}_{\text{DD}} \le 3.6 \text{ V}$	-0.3	5.5		
V <sub>IN</sub>	input voltage on FT, FTT and RST pins.	1.65 V ≤ V <sub>DD</sub> ≤2.0 V	-0.3	5.2	V	
	Input voltage on BOOT0 pin	-	0	5.5	v	
	Input voltage on TC pin	-	-0.3	V <sub>DD</sub> +0.3		
	Power dissipation at T <sub>A</sub> = 85 °C (range 6) or T <sub>A</sub> =105 °C (rage 7) $^{(3)}$	LQFP32 package	-	333		
		UFQFPN32 package	-	513		
		UFQFPN28 package	-	206		
		WLCSP25 package	-	286		
		TSSOP20 package	-	270		
		UFQFPN20 package	-	196		
P <sub>D</sub>		TSSOP14 package	-	210		
ΓD		LQFP32 package	-	83	11177	
		UFQFPN32 package	-	128		
		UFQFPN28 package	-	52		
	Power dissipation at $T_A = 125 \degree C$ (range 3) <sup>(3)</sup>	WLCSP25 package	-	71	-	
	- /	TSSOP20 package	-	67		
		UFQFPN20 package	-	49		
		TSSOP14 package	-	53		



Symbol	Parameter	Conditions	Min	Max	Unit
		Maximum power dissipation (range 6)	-40	85	
Та	Temperature range	Maximum power dissipation (range 7)	-40	105	
		Maximum power dissipation (range 3)	-40	125	°C
	Junction temperature range (range 6)	-40 °C $\leq$ T <sub>A</sub> $\leq$ 85 °	-40	105	
TJ	Junction temperature range (range 7)	-40 °C $\leq$ T <sub>A</sub> $\leq$ 105 °C	-40	125	
	Junction temperature range (range 3)	-40 °C $\leq$ T <sub>A</sub> $\leq$ 125 °C	-40	130	

1. It is recommended to power  $V_{DD}$  and  $V_{DDA}$  from the same source. A maximum difference of 300 mV between  $V_{DD}$  and  $V_{DDA}$  can be tolerated during power-up and normal operation.

2. To sustain a voltage higher than  $V_{DD}$ +0.3V, the internal pull-up/pull-down resistors must be disabled.

If T<sub>A</sub> is lower, higher P<sub>D</sub> values are allowed as long as T<sub>J</sub> does not exceed T<sub>J</sub> max (see Table 17: Thermal characteristics on page 47).

### 6.3.2 Embedded reset and power control block characteristics

The parameters given in the following table are derived from the tests performed under the ambient temperature condition summarized in *Table 18*.

Symbol	Parameter Conditions		Min	Тур	Max	Unit	
	V rice time rete	BOR detector enabled	0	-	∞		
t <sub>VDD</sub> <sup>(1)</sup>	V <sub>DD</sub> rise time rate	BOR detector disabled	0	-	1000		
'VDD`'	V fall time rate	BOR detector enabled	20	-	∞	µs/V	
	V <sub>DD</sub> fall time rate	BOR detector disabled	0	-	1000		
т (1)	Reset temporization	V <sub>DD</sub> rising, BOR enabled	-	2	3.3	ma	
T <sub>RSTTEMPO</sub> <sup>(1)</sup>		V <sub>DD</sub> rising, BOR disabled <sup>(2)</sup>	0.4	0.7	1.6	ms	
M	Power on/power down reset threshold	Falling edge	1	1.5	1.65		
V <sub>POR/PDR</sub>		Rising edge	1.3	1.5	1.65		
V	Brown-out reset threshold 0	Falling edge	1.67	1.7	1.74		
V <sub>BOR0</sub>		Rising edge	1.69	1.76	1.8	V	
V	Brown-out reset threshold 1	Falling edge	1.87	1.93	1.97	v	
V <sub>BOR1</sub>		Rising edge	1.96	2.03	2.07		
M	Brown-out reset threshold 2	Falling edge	2.22	2.30	2.35		
V <sub>BOR2</sub>		Rising edge	2.31	2.41	2.44		

### Table 19. Embedded reset and power control block characteristics



Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
M	Brown-out reset threshold 3	Falling edge	2.45	2.55	2.6	
V <sub>BOR3</sub>	Brown-out reset threshold 5	Rising edge	2.54	2.66	2.7	
M	Brown-out reset threshold 4	Falling edge	2.68	2.8	2.85	
V <sub>BOR4</sub>	Brown-out reset threshold 4	Rising edge	2.78	2.9	2.95	
V	Programmable voltage detector	Falling edge	1.8	1.85	1.88	
V <sub>PVD0</sub>	threshold 0	Rising edge	1.88	1.94	1.99	
V	PVD threshold 1	Falling edge	1.98	2.04	2.09	
V <sub>PVD1</sub>		Rising edge	2.08	2.14	2.18	
V <sub>PVD2</sub>	PVD threshold 2	Falling edge	2.20	2.24	2.28	v
		Rising edge	2.28	2.34	2.38	v
V	PVD threshold 3	Falling edge	2.39	2.44	2.48	
V <sub>PVD3</sub>		Rising edge	2.47	2.54	2.58	
V	PVD threshold 4	Falling edge	2.57	2.64	2.69	
V <sub>PVD4</sub>		Rising edge	2.68	2.74	2.79	
V	PVD threshold 5	Falling edge	2.77	2.83	2.88	
V <sub>PVD5</sub>		Rising edge	2.87	2.94	2.99	
N/		Falling edge	2.97	3.05	3.09	
V <sub>PVD6</sub>	PVD threshold 6	Rising edge	3.08	3.15	3.20	
		BOR0 threshold	-	40	-	
V <sub>hyst</sub>	Hysteresis voltage	All BOR and PVD thresholds excepting BOR0	-	100	-	mV

Table 19. Embedded reset and power control block characteristics (continued)

1. Guaranteed by characterization results, not tested in production.

2. Valid for device version without BOR at power up. Please see option "D" in Ordering information scheme for more details.



### Embedded internal reference voltage 6.3.3

The parameters given in Table 21 are based on characterization results, unless otherwise specified.

Table 20. Embedde	ed internal reference voltage	calibration values
Calibration value name	Description	Memory address
	Raw data acquired at temperature of 25°C	0x1FF8 0078 - 0x1FF8 0079

Table 21. Embedded internal reference voltage <sup>(1)</sup>								
Symbol	Parameter	Conditions	Min	Тур	Max	Unit		
V <sub>REFINT out</sub> <sup>(2)</sup>	Internal reference voltage	– 40 °C < T <sub>J</sub> < +125 °C	1.202	1.224	1.242	V		
T <sub>VREFINT</sub>	Internal reference startup time	-	-	2	3	ms		
V <sub>VREF_MEAS</sub>	V <sub>DDA</sub> voltage during V <sub>REFINT</sub> factory measure	-	2.99	3	3.01	V		
A <sub>VREF_MEAS</sub>	Accuracy of factory-measured V <sub>REFINT</sub> value <sup>(3)</sup>		-	-	±5	mV		
т (4)		–40 °C < T <sub>J</sub> < +125 °C	-	25	100	ppm/°C		
T <sub>Coeff</sub> <sup>(4)</sup>	Temperature coefficient	0 °C < T <sub>J</sub> < +50 °C	-	-	20	ppin/ C		
A <sub>Coeff</sub> <sup>(4)</sup>	Long-term stability	1000 hours, T= 25 °C	-	-	1000	ppm		
V <sub>DDCoeff</sub> <sup>(4)</sup>	Voltage coefficient	3.0 V < V <sub>DDA</sub> < 3.6 V	-	-	2000	ppm/V		
T <sub>S_vrefint</sub> <sup>(4)(5)</sup>	ADC sampling time when reading the internal reference voltage	-	5	10	-	μs		
T <sub>ADC_BUF</sub> <sup>(4)</sup>	Startup time of reference voltage buffer for ADC	-	-	-	10	μs		
I <sub>BUF_ADC</sub> <sup>(4)</sup>	Consumption of reference voltage buffer for ADC	-	-	13.5	25	μA		
I <sub>VREF_OUT</sub> <sup>(4)</sup>	VREF_OUT output current <sup>(6)</sup>	-	-	-	1	μA		
C <sub>VREF_OUT</sub> <sup>(4)</sup>	VREF_OUT output load	-	-	-	50	pF		
I <sub>LPBUF</sub> <sup>(4)</sup>	Consumption of reference voltage buffer for VREF_OUT and COMP	-	-	730	1200	nA		
V <sub>REFINT_DIV1</sub> <sup>(4)</sup>	1/4 reference voltage	-	24	25	26	0.1		
V <sub>REFINT_DIV2</sub> <sup>(4)</sup>	1/2 reference voltage	-	49	50	51	% V <sub>REFINT</sub>		
V <sub>REFINT_DIV3</sub> <sup>(4)</sup>	3/4 reference voltage	-	74	75	76			

### Table 21, Embedded internal reference voltage<sup>(1)</sup>

V<sub>DDA</sub>= 3 V

Refer to Table 33: Peripheral current consumption in Stop and Standby mode for the value of the internal reference current consumption ( $I_{REFINT}$ ). 1.

2. Guaranteed by test in production.

3. The internal V<sub>REF</sub> value is individually measured in production and stored in dedicated EEPROM bytes.



- 4. Guaranteed by design, not tested in production.
- 5. Shortest sampling time can be determined in the application by multiple iterations.
- 6. To guarantee less than 1% VREF\_OUT deviation.

### 6.3.4 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code. The current consumption is measured as described in *Figure 13: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to Dhrystone 2.1 code if not specified otherwise.

The current consumption values are derived from the tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18: General operating conditions* unless otherwise specified.

The MCU is placed under the following conditions:

- All I/O pins are configured in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time and prefetch is adjusted depending on f<sub>HCLK</sub> frequency and voltage range to provide the best CPU performance unless otherwise specified.
- When the peripherals are enabled f<sub>APB1</sub> = f<sub>APB2</sub> = f<sub>APB</sub>
- When PLL is ON, the PLL inputs are equal to HSI = 16 MHz (if internal clock is used) or HSE = 16 MHz (if HSE bypass mode is used)
- The HSE user clock is applied to CK\_IN. It follows the characteristic specified in Table 35: High-speed external user clock characteristics
- For maximum current consumption  $V_{DD} = V_{DDA} = 3.6$  V is applied to all supply pins
- For typical current consumption V<sub>DD</sub> = V<sub>DDA</sub> = 3.0 V is applied to all supply pins if not specified otherwise



Symbol	Parameter	Co	nditions	f <sub>HCLK</sub>	Тур	Max <sup>(1)</sup>	Unit
				1 MHz	140	180	
			Range 3, V <sub>CORE</sub> =1.2 V VOS[1:0]=11	2 MHz	245	290	μA
				4 MHz	460	540	
	16 MHz includ	f <sub>HSE</sub> = f <sub>HCLK</sub> up to		4 MHz	0.56	0.65	
		16 MHz included, f <sub>HSE</sub> = f <sub>HCLK</sub> /2 above	Range 2, V <sub>CORE</sub> =1.5 V, VOS[1:0]=10,	8 MHz	1.1	1.3	
Supply I <sub>DD</sub> current in	16 MHz (PLL ON) <sup>(2)</sup>		16 MHz	2.1	2.4	mA	
			8 MHz	1.3	1.6		
(Run from	Run Run mode,		Range 1, V <sub>CORE</sub> =1.8 V, VOS[1:0]=01	16 MHz	2.6	3	
Flash)	executed			32 MHz	5.3	6.5	
	from Flash	MSI clock	Range 3, V <sub>CORE</sub> =1.2 V, VOS[1:0]=11	65 kHz	34.5	54	
	MSI cl			524 kHz	86	120	μA
			4.2 MHz	505	560		
			Range 2, V <sub>CORE</sub> =1.5 V, VOS[1:0]=10,	16 MHz	2.2	2.6	
	HSI clock		Range 1, V <sub>CORE</sub> =1.8 V, VOS[1:0]=01	32 MHz	5.4	5.9	mA

Table 22 Current consumption in Run mode	, code with data processing running from Flash
Table 22. Outfent consumption in Run mode	, code with data processing running nom riash

2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

Table 23. Current consumption in Run mode vs code type,
code with data processing running from Flash

Symbol	Parameter		Conditions				Unit
				Dhrystone		460	
			CoreMark		440	μA	
		Range 3, V <sub>CORE</sub> =1.2 V, VOS[1:0]=11	Fibonacci	4 MHz	330		
			while(1)		305		
l <sub>DD</sub> (Run		f <sub>HSE</sub> = f <sub>HCLK</sub> up to 16 MHz included,		while(1), prefetch OFF		320	
from code f <sub>HSE</sub> = f <sub>HO</sub>	f <sub>HSE</sub> = f <sub>HCLK</sub> /2 above 16 MHz (PLL ON) <sup>(1)</sup>	Range 1, VOS[1:0]=01,	Dhrystone	32 MHz	5.4		
			CoreMark		4.9	mA	
			Fibonacci		5		
		V <sub>CORE</sub> =1.8 V	while(1)		4.35		
			while(1), prefetch OFF		3.7		

1. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).



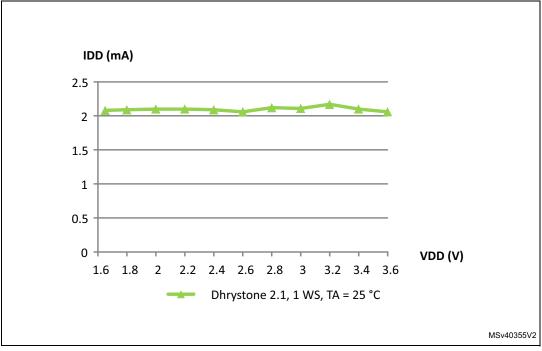
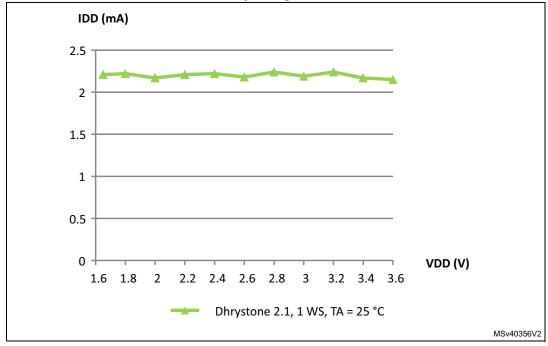


Figure 14. I<sub>DD</sub> vs V<sub>DD</sub>, at T<sub>A</sub>= 25 °C, Run mode, code running from Flash memory, Range 2, 16 MHz HSE, 1WS

Figure 15.  $I_{DD}$  vs  $V_{DD}$ , at  $T_A$ = 25 °C, Run mode, code running from Flash memory, Range 2, HSI16, 1WS





Symbol	Parameter	Conc	litions	f <sub>HCLK</sub>	Тур	Max <sup>(1)</sup>	Unit
Supply current in I <sub>DD</sub> (Run Run mode, code			Range 3,	1 MHz	115	140	
		V <sub>CORE</sub> =1.2 V,	2 MHz	205	240	μA	
		VOS[1:0]=11	4 MHz	385	420		
	f <sub>HSE</sub> = f <sub>HCLK</sub> up to 16 MHz, included	Range 2,	4 MHz	0.48	0.55		
	$f_{HSE} = f_{HCLK}/2$ above	V <sub>CORE</sub> =1.5 ,V,	8 MHz	0.935	1.1	mA	
	16 MHz (PLL ON) <sup>(2)</sup>	VOS[1:0]=10	16 MHz	1.8	2		
	(,	Range 1, V <sub>CORE</sub> =1.8 V,	8 MHz	1.1	1.4		
			16 MHz	2.1	2.5		
from RAM)	executed from RAM, Flash		VOS[1:0]=01	32 MHz	4.5	4.9	
	switched OFF	MSI clock	Range 3, V <sub>CORE</sub> =1.2 V, VOS[1:0]=11	65 kHz	22	38	μΑ
				524 kHz	67	91	
				4.2 MHz	415	450	
	HSI16 clock source	Range 2, V <sub>CORE</sub> =1.5 V, VOS[1:0]=10	16 MHz	1.95	2.2		
	(16 MHz)	Range 1, V <sub>CORE</sub> =1.8 V, VOS[1:0]=01	32 MHz	4.7	5.2	mA	

2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

# Table 25. Current consumption in Run mode vs code type,code with data processing running from RAM<sup>(1)</sup>

Symbol	Parameter		f <sub>HCLK</sub>	Тур	Unit		
				Dhrystone		385	
			Range 3,	CoreMark		_(3)	
$ \begin{array}{c} I_{DD} \ (Run \\ from \\ RAM) \end{array} \begin{array}{c} Supply \ current \ in \\ Run \ mode, \ code \\ executed \ from \\ RAM, \ Flash \\ switched \ OFF \end{array} \begin{array}{c} f_{HSE} = f_{HCLK} \ up \ to \ 16 \\ MHz, \ included, \\ f_{HSE} = f_{HCLK}/2 \ above \\ 16 \ MHz \ (PLL \ ON)^{(2)} \end{array} $	V <sub>CORE</sub> =1.2 V, VOS[1:0]=11	Fibonacci	4 MHz	350	μA		
			while(1)	1	340		
	$f_{HSE} = f_{HCLK}/2$ above		Dhrystone		4.5		
		Range 1, V <sub>CORE</sub> =1.8 V,	CoreMark	- 32 MHz	_(3)	mA	
		VOS[1:0]=01	Fibonacci		4.2	ША	
			while(1)		3		

1. Guaranteed by characterization results, not tested in production, unless otherwise specified.

2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).

3. CoreMark code is unable to run from RAM since the RAM size is only 2 Kbytes.



### **Electrical characteristics**

Symbol	Parameter	Conc	litions	f <sub>HCLK</sub>	Тур	Max <sup>(1)</sup>	Unit	
			Range 3,	1 MHz	36.5	70		
			V <sub>CORE</sub> =1.2 V,	2 MHz	58	95		
			VOS[1:0]=11	4 MHz	100	150		
		f <sub>HSE</sub> = f <sub>HCLK</sub> up to 16 MHz included,	Range 2,	4 MHz	125	170		
		$f_{HSE} = f_{HCLK}/2$	V <sub>CORE</sub> =1.5 V,	8 MHz	230	300		
		above 16 MHz (PLL ON) <sup>(2)</sup>	VOS[1:0]=10	16 MHz	450	540		
		(1 22 011)	Range 1,	8 MHz	275	350		
	Supply current		V <sub>CORE</sub> =1.8 V,	16 MHz	555	650		
	in Sleep mode, Flash		VOS[1:0]=01	32 MHz	1350	1600		
	OFF		Range 3,	65 kHz	15.5	32		
		MSI clock	V <sub>CORE</sub> =1.2 V,	524 kHz	26.5	55		
			VOS[1:0]=11	4.2 MHz	115	160		
		HSI16 clock source (16 MHz)	HSI16 clock source	Range 2, V <sub>CORE</sub> =1.5 V, VOS[1:0]=10	16 MHz	585	670	
			Range 1, V <sub>CORE</sub> =1.8 V, VOS[1:0]=01	32 MHz	1500	1700	<b>-</b>	
I <sub>DD</sub> (Sleep)		$f_{HSE} = f_{HCLK}$ up to 16 MHz included, $f_{HSE} = f_{HCLK}/2$ above 16 MHz (PLL ON) <sup>(2)</sup>	Range 3, V <sub>CORE</sub> =1.2 V, VOS[1:0]=11	1 MHz	49	88	μA	
				2 MHz	69	120	1	
				4 MHz	115	190		
			Range 2, <sub>CORE</sub> =1.5 V, VOS[1:0]=10	4 MHz	135	200		
				8 MHz	240	340		
				16 MHz	460	650		
			Range 1, V <sub>CORE</sub> =1.8 V,	8 MHz	290	400		
	Supply current			16 MHz	565	750	-	
	in Sleep mode, Flash		VOS[1:0]=01	32 MHz	1350	1900		
	ON		Range 3,	65 kHz	26.5	46		
		MSI clock	V <sub>CORE</sub> =1.2 V,	524 kHz	38.5	70	-	
			VOS[1:0]=11	4.2 MHz	125	190		
		HSI16 clock source (16 MHz)	Range 2, V <sub>CORE</sub> =1.5 V, VOS[1:0]=10	16 MHz	600	760		
				32 MHz	1500	1850		

1. Guaranteed by characterization results at 125 °C, not tested in production, unless otherwise specified.

2. Oscillator bypassed (HSEBYP = 1 in RCC\_CR register).



Symbol	Parameter	Conditions				Max <sup>(1)</sup>	Unit
				$T_A$ = -40 °C to 25 °C	5.7	8.1	
			MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 85 °C	6.5	9	
				T <sub>A</sub> = 105 °C	8	13	
		All		T <sub>A</sub> = 125 °C	11.5	22	
		peripherals OFF, code		$T_A = -40 \degree C$ to 25 $\degree C$	8.7	11	
		executed	MSI clock, 65 kHz	T <sub>A</sub> = 85 °C	9.5	12	
		from RAM, Flash	f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 105 °C	11	15	
		switched		T <sub>A</sub> = 125 °C	15	24	
		OFF, V <sub>DD</sub> from 1.65 V		$T_A = -40 \ ^\circ C$ to 25 $^\circ C$	17	19	
Querela		to 3.6 V		T <sub>A</sub> = 55 °C	17	19.5	- μΑ
			MSI clock, 131 kHz f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85 °C	17.5	20	
	Supply			T <sub>A</sub> = 105 °C	19	22	
I <sub>DD</sub>	current in			T <sub>A</sub> = 125 °C	22.5	31	
(LP Run)	Low-power	All peripherals OFF, code	MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	$T_A$ = -40 °C to 25 °C	18	22	
	Turr mode			T <sub>A</sub> = 85 °C	20	24	
				T <sub>A</sub> = 105 °C	22	27	
				T <sub>A</sub> = 125 °C	26.5	37	
			MSI clock, 65 kHz	$T_A$ = -40 °C to 25 °C	22	25	
				T <sub>A</sub> = 85 °C	24	27	
		executed from Flash,	f <sub>HCLK</sub> = 65 kHz	T <sub>A</sub> = 105 °C	26	30	
		V <sub>DD</sub> from		T <sub>A</sub> = 125 °C	30.5	39	
		1.65 V to 3.6 V		$T_A = -40 \ ^\circ C$ to 25 $^\circ C$	32	34	
				T <sub>A</sub> = 55 °C	32.5	35	
			MSI clock, 131 kHz f <sub>HCLK</sub> = 131 kHz	T <sub>A</sub> = 85 °C	34	37	
			HOLK	T <sub>A</sub> = 105 °C	36	39	
				T <sub>A</sub> = 125 °C	40	47	



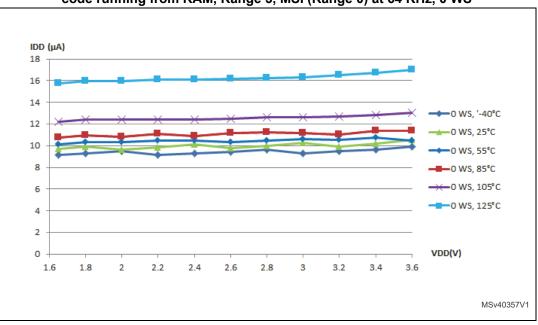


Figure 16. I<sub>DD</sub> vs V<sub>DD</sub>, at T<sub>A</sub>= -40/25/55/ 85/105/125 °C, Low-power run mode, code running from RAM, Range 3, MSI (Range 0) at 64 KHz, 0 WS

Table 28. Current consum	ption in Low-power Sleep mode

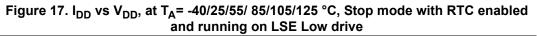
Symbol	Parameter	Conditions				Max <sup>(1)</sup>	Unit
			MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz Flash OFF	$T_A = -40 \ ^\circ C$ to 25 $^\circ C$	2.5 <sup>(2)</sup>	-	
				$T_A$ = -40 °C to 25 °C	13	19	
			MSI clock, 65 kHz f <sub>HCLK</sub> = 32 kHz	T <sub>A</sub> = 85 °C	15.5	20	
		Flash ON	T <sub>A</sub> = 105 °C	17.5	22		
		All peripherals OFF, V <sub>DD</sub> from 1.65 V to 3.6 V		T <sub>A</sub> = 125 °C	21	29	
	Supply current in		MSI clock, 65 kHz f <sub>HCLK</sub> = 65 kHz, Flash ON	T <sub>A</sub> = -40 °C to 25 °C	13.5	19	
I <sub>DD</sub> (LP Sleep)	Low-power			T <sub>A</sub> = 85 °C	16	20	μA
	sleep mode			T <sub>A</sub> = 105 °C	18	22	
				T <sub>A</sub> = 125 °C	21.5	29	
				$T_A$ = -40 °C to 25 °C	15.5	21	
			MSI clock, 131 kHz	T <sub>A</sub> = 55 °C	17	22	
			f <sub>HCLK</sub> = 131 kHz,	T <sub>A</sub> = 85 °C	18	23	
			Flash ON	T <sub>A</sub> = 105 °C	19.5	24	
				T <sub>A</sub> = 125 °C	23.5	31	

 As the CPU is in Sleep mode, the difference between the current consumption with Flash memory ON and OFF (nearly 12 μA) is the same whatever the clock frequency.



	Bereventer		•	Max <sup>(1)</sup>	11
Symbol	Parameter	Conditions	Тур	Max.,	Unit
		$T_A = -40^{\circ}C$ to $25^{\circ}C$	0.34	0.99	
		T <sub>A</sub> = 55°C	0.43	1.9	
I <sub>DD</sub> (Stop)	Supply current in Stop mode	T <sub>A</sub> = 85°C	0.94	4.2	μA
		T <sub>A</sub> = 105°C	2.0	9	
		T <sub>A</sub> = 125°C	4.9	19	

Table 29. Typical and maximum current consumptions in Stop mode



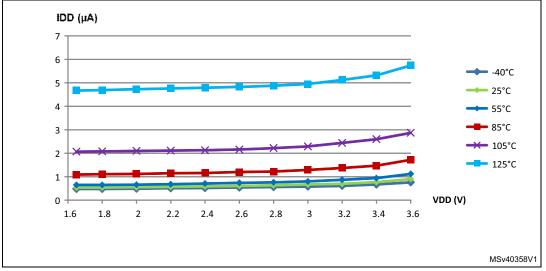
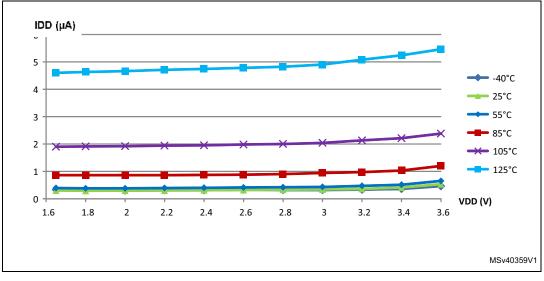


Figure 18. I<sub>DD</sub> vs V<sub>DD</sub>, at T<sub>A</sub>= -40/25/55/85/105/125 °C, Stop mode with RTC disabled, all clocks OFF





Symbol	Parameter	Conditions		Тур	Max <sup>(1)</sup>	Unit
IDD			$T_A$ = -40 °C to 25 °C	0.8	1.6	
			T <sub>A</sub> = 55 °C	0.9	1.8	
		Independent watchdog and LSI enabled	T <sub>A</sub> = 85 °C	1	2	
			T <sub>A</sub> = 105 °C	1.25	3	
	Supply current in Standby		T <sub>A</sub> = 125 °C	2	7	
(Standby)	mode		$T_A$ = -40 °C to 25 °C	0.23	0.6	μA
			T <sub>A</sub> = 55 °C	0.25	0.7	
		Independent watchdog and LSI OFF	T <sub>A</sub> = 85 °C	0.36	1	
			T <sub>A</sub> = 105 °C	0.62	1.7	
			T <sub>A</sub> = 125 °C	1.35	5	

Table 30. Typical and maximum current consumptions in	Standby mode
---	--------------

Symbol	parameter	System frequency	Current consumption during wakeup	Unit
		HSI	1	
		HSI/4	0,7	
I <sub>DD</sub> (WU from Stop)	Supply current during wakeup from Stop mode	MSI 4,2 MHz	0,7	
		MSI 1,05 MHz	0,4	
		MSI 65 KHz	0,1	mA
I <sub>DD</sub> (Reset)	Reset pin pulled down	-	0,21	
I <sub>DD</sub> (Power Up)	BOR ON	-	0,23	
I <sub>DD</sub> (WU from	With Fast wakeup set	MSI 2,1 MHz	0,5	
StandBy)	With Fast wakeup disabled	MSI 2,1 MHz	0,12	

### **On-chip peripheral current consumption**

The current consumption of the on-chip peripherals is given in the following tables. The MCU is placed under the following conditions:

- all I/O pins are in input mode with a static value at  $V_{DD}$  or  $V_{SS}$  (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
  - with all peripherals clocked OFF
  - with only one peripheral clocked ON



		Typical	consumption, V	/ <sub>DD</sub> = 3.0 V, T <sub>A</sub> =	25 °C		
Peripheral		Range 1, V <sub>CORE</sub> =1.8 V VOS[1:0] = 01	Range 2, V <sub>CORE</sub> =1.5 V VOS[1:0] = 10	Range 3, V <sub>CORE</sub> =1.2 V VOS[1:0] = 11	Low-power sleep and run	Unit	
	WWDG	2.5	2	1.6	2		
	LPUART1	8.3	7.2	5.4	7.2		
APB1	I2C1	11	8.2	6.8	8.9	µA/MHz	
	LPTIM1	14	11	8.7	11	(f <sub>HCLK</sub> )	
	TIM2	10.5	8.5	6.4	8.5		
	USART2	8.5	6.8	5.4	7.1		
	ADC1 <sup>(2)</sup>	5.0	3.9	3.3	4		
	SPI1	4.5	3.5	2.9	3.6	µA/MHz	
APB2	TIM21	6.8	6.1	4.5	5.6		
	DBGMCU	1.7	1.7	1.1	1.4	(f <sub>HCLK</sub> )	
	SYSCFG/ COMP	2.5	2.4	1.6	2.3		
Cortex-	GPIOA	7.6	6.3	4.9	6.5		
M0+ core	GPIOB	5.1	4.1	3.2	4	µA/MHz (f <sub>HCLK</sub> )	
I/O port	GPIOC	1.1	0.7	0.6	0.8	('HCLK)	
	CRC	1.5	1.1	1	1.2		
	FLASH <sup>(3)</sup>	10	8.5	7	8.5		
AHB	DMA1	5.3	4.2	3.5	4.8	µA/MHz (f <sub>HCLK</sub> )	
All enabled		96	80	62	88	VHULK)	
PWR		2.1	1.9	1.4	1.8	µA/MHz (f <sub>HCLK</sub> )	

Table 32. Periph	eral current consumption in run or Sleep mode <sup>(1)</sup>	
------------------	--	--

 Data based on differential I<sub>DD</sub> measurement between all peripherals OFF an one peripheral with clock enabled, in the following conditions: f<sub>HCLK</sub> = 32 MHz (range 1), f<sub>HCLK</sub> = 16 MHz (range 2), f<sub>HCLK</sub> = 4 MHz (range 3), f<sub>HCLK</sub> = 64kHz (Low-power run/sleep), f<sub>APB1</sub> = f<sub>HCLK</sub>, f<sub>APB2</sub> = f<sub>HCLK</sub>, default prescaler value for each peripheral. The CPU is in Sleep mode in both cases. No I/O pins toggling. Not tested in production.

2. HSI oscillator is OFF for this measure.

 These values correspond to the Flash memory dynamic current consumption. The Flash memory static consumption (Flash memory ON) equals 12 µA and does not depend on the frequency. The Flash memory consumption is already taken into account in all the supply current consumption tables (Flash memory ON cases).



Symbol	Peripheral	Typical consumption, T <sub>A</sub> = 25 °C		Unit
Symbol	Peripheral	V <sub>DD</sub> =1.8 V	V <sub>DD</sub> =3.0 V	Unit
I <sub>DD(PVD / BOR)</sub>	-	0.6	1	
I <sub>REFINT</sub>	-	1.25	1.3	
-	LSE Low drive	0.11	0.16	
-	LPTIM1, Input 100 Hz	0.01	0.02	μΑ
-	LPTIM1, Input 1 MHz	8	9	
-	LPUART1	0.025	0.027	
-	RTC	0.1	0.19	1

Table 33. Peripheral current consumption in Stop and Standby mode

### 6.3.5 Wakeup time from low-power mode

The wakeup times given in the following table are measured with the MSI or HSI16 RC oscillator. The clock source used to wake up the device depends on the current operating mode:

- Sleep mode: the clock source is the clock that was set before entering Sleep mode
- Stop mode: the clock source is either the MSI oscillator in the range configured before entering Stop mode, the HSI16 or HSI16/4.
- Standby mode: the clock source is the MSI oscillator running at 2.1 MHz

All timings are derived from tests performed under ambient temperature and V<sub>DD</sub> supply voltage conditions summarized in *Table 18*.

Symbol	Parameter	Conditions	Тур	Max	Unit
t <sub>WUSLEEP</sub>	Wakeup from Sleep mode	f <sub>HCLK</sub> = 32 MHz	7	8	
	Wakeup from Low-power sleep mode,	f <sub>HCLK</sub> = 262 kHz Flash enabled	7	8	CPU cvcles
	f <sub>HCLK</sub> = 262 kHz	f <sub>HCLK</sub> = 262 kHz Flash switched OFF	9	10	,

Table 34. Low-power mode wakeup timings



Symbol	Parameter	Conditions	Тур	Max	Unit
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz	5.1	8	
Symbol	Wakeup from Stop mode, regulator in Run mode	f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	5.1	7	
		$f_{HCLK} = f_{HSI}/4 = 4 \text{ MHz}$	8.1	11	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 1	5	8	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 2	5	8	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz Voltage range 3	5	8	3 3 3 5 µs 0 0
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 2.1 MHz	7.4		
	Wakeup from Stop mode, regulator in low-power mode	f <sub>HCLK</sub> = f <sub>MSI</sub> = 1.05 MHz	14	23	-
t <sub>WUSTOP</sub>		f <sub>HCLK</sub> = f <sub>MSI</sub> = 524 kHz	28	38	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 262 kHz	51	65	μs
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 131 kHz	99	120	
		f <sub>HCLK</sub> = f <sub>MSI</sub> = 65 kHz	196	260	
		f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	5.1	7	
		$f_{HCLK} = f_{HSI}/4 = 4 \text{ MHz}$	8.2	11	
	Wakeup from Stop mode, regulator in low-power mode, HSI kept running in Stop mode	f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	3.25	-	
	Wakeup from Stop mode, regulator in	f <sub>HCLK</sub> = f <sub>HSI</sub> = 16 MHz	4.9	7	1
	low-power mode, code running from	$f_{HCLK} = f_{HSI}/4 = 4 \text{ MHz}$	7.9	10	
	RAM	f <sub>HCLK</sub> = f <sub>MSI</sub> = 4.2 MHz	4.8	8	
tumores	Wakeup from Standby mode FWU bit = 1	f <sub>HCLK</sub> = f <sub>MSI</sub> = 2.1 MHz	65	130	
•WUSTDBY	Wakeup from Standby mode FWU bit = 0	f <sub>HCLK</sub> = f <sub>MSI</sub> = 2.1 MHz	2.2	3	ms

Table 34. Low-power mode wakeup timings (continued)



### 6.3.6 External clock source characteristics

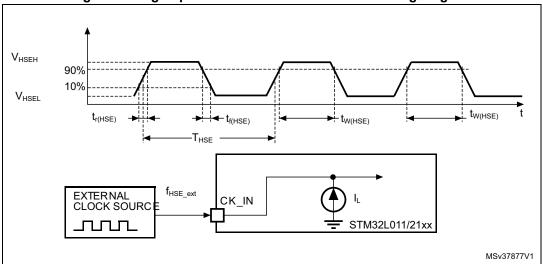
### High-speed external user clock generated from an external source

In bypass mode the input pin is a standard GPIO. The external clock signal has to respect the I/O characteristics in *Section 6.3.12*. However, the recommended clock input waveform is shown in *Figure 19*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f	User external clock source	CSS is ON or PLL is used	1	8	32	MHz
f <sub>HSE_ext</sub>	frequency	CSS is OFF, PLL not used	0	8	32	MHz
V <sub>HSEH</sub>	CK_IN input pin high level voltage		$0.7V_{DD}$	-	V <sub>DD</sub>	V
V <sub>HSEL</sub>	CK_IN input pin low level voltage		$V_{SS}$	-	$0.3V_{\text{DD}}$	v
t <sub>w(HSE)</sub> t <sub>w(HSE)</sub>	CK_IN high or low time	_	12	-	-	ns
t <sub>r(HSE)</sub> t <sub>f(HSE)</sub>	CK_IN rise or fall time		-	-	20	115
C <sub>in(HSE)</sub>	CK_IN input capacitance		-	2.6	-	pF
DuCy <sub>(HSE)</sub>	Duty cycle		45	-	55	%
١L	CK_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	_	±1	μA

Table 35. High-speed external user clock characteristics<sup>(1)</sup>

1. Guaranteed by design, not tested in production.



### Figure 19. High-speed external clock source AC timing diagram



### Low-speed external user clock generated from an external source

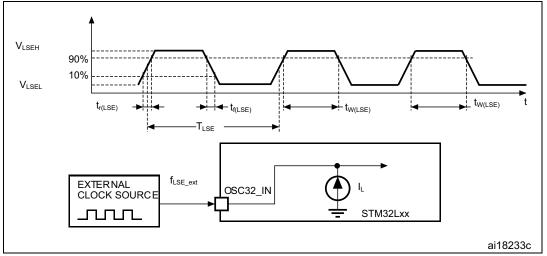
The characteristics given in the following table result from tests performed using a lowspeed external clock source, and under ambient temperature and supply voltage conditions summarized in *Table 18*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User external clock source frequency		-	32.768	1000	kHz
V <sub>LSEH</sub>	OSC32_IN input pin high level voltage		0.7V <sub>DD</sub>	-	V <sub>DD</sub>	v
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage	_	V <sub>SS</sub>	-	0.3V <sub>DD</sub>	v
t <sub>w(LSE)</sub> t <sub>w(LSE)</sub>	OSC32_IN high or low time		465	-	-	ns
t <sub>r(LSE)</sub> t <sub>f(LSE)</sub>	OSC32_IN rise or fall time		-	-	10	115
C <sub>IN(LSE)</sub>	OSC32_IN input capacitance	-	-	0.6	-	pF
DuCy <sub>(LSE)</sub>	Duty cycle	-	45	-	55	%
١L	OSC32_IN Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	±1	μA

 Table 36. Low-speed external user clock characteristics<sup>(1)</sup>

1. Guaranteed by design, not tested in production





### Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 37*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization



time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions <sup>(2)</sup>		Тур	Max	Unit		
f <sub>LSE</sub>	LSE oscillator frequency		-	32.768	-	kHz		
G <sub>m</sub>		LSEDRV[1:0]=00 lower driving capability	-	-	0.5			
		Maximum critical crystal	LSEDRV[1:0]= 01 medium low driving capability	-	-	0.75		
	transconductance	LSEDRV[1:0] = 10 medium high driving capability	-	-	1.7	μΑ/V		
		LSEDRV[1:0]=11 higher driving capability	-	-	2.7			
t <sub>SU(LSE)</sub> <sup>(3)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	s		

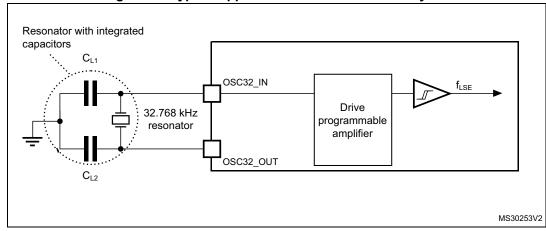
Table 37. LSE	oscillator	characteristics <sup>(1)</sup>
---------------	------------	--------------------------------

1. Guaranteed by design, not tested in production.

2. Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".

 Guaranteed by characterization results, not tested in production. t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer. To increase speed, address a lower-drive quartz with a high- driver mode.

# *Note:* For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website <u>http://www.st.com</u>.



### Figure 21. Typical application with a 32.768 kHz crystal

*Note:* An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.



### 6.3.7 Internal clock source characteristics

The parameters given in *Table 38* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18*.

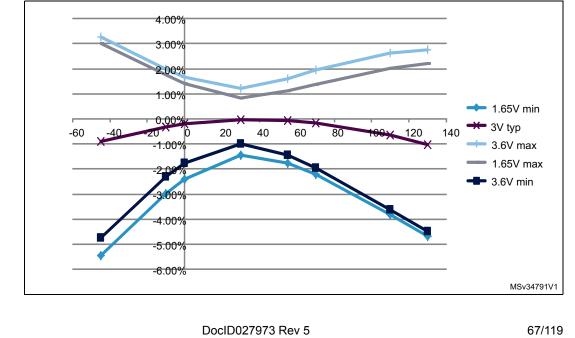
### High-speed internal 16 MHz (HSI16) RC oscillator

Symbol	Parameter	Conditions		Тур	Max	Unit
f <sub>HSI16</sub>	Frequency	V <sub>DD</sub> = 3.0 V	-	16	-	MHz
TRIM <sup>(1)(2)</sup>	HSI16 user-	Trimming code is not a multiple of 16	-	±0.4	0.7	%
TRIM	trimmed resolution	Trimming code is a multiple of 16	-	-	±1.5	%
		V <sub>DDA</sub> = 3.0 V, T <sub>A</sub> = 25 °C	-1 <sup>(3)</sup>	-	1 <sup>(3)</sup>	%
	Accuracy of the factory-calibrated HSI16 oscillator	$V_{DDA}$ = 3.0 V, $T_A$ = 0 to 55 °C	-1.5	-	1.5	%
ACC		$V_{DDA}$ = 3.0 V, $T_A$ = -10 to 70 °C	-2	-	2	%
ACC <sub>HSI16</sub>		$V_{DDA}$ = 3.0 V, $T_A$ = -10 to 85 °C	-2.5	-	2	%
		$V_{DDA}$ = 3.0 V, $T_A$ = -10 to 105 °C	-4	-	2	%
		V <sub>DDA</sub> = 1.65 V to 3.6 V T <sub>A</sub> = -40 to 125 °C	-5.45	-	3.25	%
t <sub>SU(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator startup time	-	-	3.7	6	μs
I <sub>DD(HSI16)</sub> <sup>(2)</sup>	HSI16 oscillator power consumption	-	-	100	140	μA

1. The trimming step differs depending on the trimming code. It is usually negative on the codes which are multiples of 16 (0x00, 0x10, 0x20, 0x30...0xE0).

2. Guaranteed by characterization results, not tested in production.

3. Guaranteed by test in production.



### Figure 22. HSI16 minimum and maximum value versus temperature

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### Low-speed internal (LSI) RC oscillator

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>LSI</sub> <sup>(1)</sup>	LSI frequency	26	38	56	kHz
D <sub>LSI</sub> <sup>(2)</sup>	LSI oscillator frequency drift $0^{\circ}C \le T_{A} \le 85^{\circ}C$	-10	-	4	%
t <sub>su(LSI)</sub> <sup>(3)</sup>	LSI oscillator startup time	-	-	200	μs
I <sub>DD(LSI)</sub> <sup>(3)</sup>	LSI oscillator power consumption	-	400	510	nA

1. Guaranteed by test in production.

2. This is a deviation for an individual part, once the initial frequency has been measured.

3. Guaranteed by design, not tested in production.

### Multi-speed internal (MSI) RC oscillator

Symbol	Parameter	Condition	Тур	Max	Unit
Symbol	Falameter	Condition	тур	IVIAX	Unit
		MSI range 0	65.5	-	
		MSI range 1	131	-	kHz
		MSI range 2	262	-	KI IZ
f <sub>MSI</sub>	Frequency after factory calibration, done at $V_{DD}$ = 3.3 V and T <sub>A</sub> = 25 °C	MSI range 3	524	- - - - - - - - - - - -	
		MSI range 4	1.05	-	
		MSI range 5	2.1 -	MHz	
		MSI range 6	4.2	-	
ACC <sub>MSI</sub>	Frequency error after factory calibration	-	±0.5	-	%
D <sub>TEMP(MSI)</sub> <sup>(1)</sup>	MSI oscillator frequency drift $0 ^\circ\text{C} \leq \text{T}_{\text{A}} \leq 85 ^\circ\text{C}$	-	±3	-	%
D <sub>VOLT(MSI)</sub> <sup>(1)</sup>	MSI oscillator frequency drift 1.65 V $\leq$ V <sub>DD</sub> $\leq$ 3.6 V, T <sub>A</sub> = 25 °C	-	-	2.5	%/V
		MSI range 0	0.75	-	
		MSI range 1	1	-	
		MSI range 2	1.5	-	
I <sub>DD(MSI)</sub> <sup>(2)</sup>	MSI oscillator power consumption	MSI range 3	2.5	-	μA
		MSI range 4	4.5	-	
		MSI range 5	8	-	-
		MSI range 6	15	-	

### Table 40. MSI oscillator characteristics



Symbol	Parameter	Condition	Тур	Мах	Unit
		MSI range 0	30	-	
		MSI range 1 20	20	-	
		MSI range 2	15	-	
		MSI range 3	10	-	
tourion	MSI oscillator startup time	MSI range 4	6	-	μs
t <sub>SU(MSI)</sub>		MSI range 5	5		μο
		MSI range 6, Voltage range 1 and 2	3.5		
		MSI range 6, Voltage range 3	5	-	
		MSI range 0		40	
		MSI range 1			
		MSI range 2	-	10	
		MSI range 3	-	-	
t <sub>STAB(MSI)</sub> <sup>(2)</sup>	MSI oscillator stabilization time	MSI range 4	-	2.5	μs
'STAB(MSI)		MSI range 5	-	2	μο
		MSI range 6, Voltage range 1 and 2	-	2	
		MSI range 3, Voltage range 3	-	3	
f	MSI oscillator frequency overshoot	Any range to range 5	-	4	MHz
f <sub>OVER(MSI)</sub>		Any range to range 6	-	6	IVIITZ

1. This is a deviation for an individual part, once the initial frequency has been measured.

2. Guaranteed by characterization results, not tested in production.

### 6.3.8 PLL characteristics

The parameters given in *Table 41* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18*.

Table 41. PLL characteristics
-------------------------------

Symbol	Parameter		Unit		
Symbol		Min Typ Max <sup>(1)</sup>	onit		
f <sub>PLL_IN</sub>	PLL input clock <sup>(2)</sup>	2	-	24	MHz
	PLL input clock duty cycle	45	-	55	%



Symbol	P tu				
	Parameter	Min	Тур	Max <sup>(1)</sup>	Unit
f <sub>PLL_OUT</sub>	PLL output clock	2	-	32	MHz
t <sub>LOCK</sub>	PLL input = 16 MHz PLL VCO = 96 MHz	-	115	160	μs
Jitter	Cycle-to-cycle jitter	-		±600	ps
I <sub>DDA</sub> (PLL)	Current consumption on V <sub>DDA</sub>	-	220	450	
I <sub>DD</sub> (PLL)	Current consumption on V <sub>DD</sub>	-	120	150	μA

Table 41. PLL characteristics (continued)

1. Guaranteed by characterization results, not tested in production.

2. Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by  $f_{PLL_OUT}$ .

### 6.3.9 Memory characteristics

### **RAM** memory

### Table 42. RAM and hardware registers

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
VRM	Data retention mode <sup>(1)</sup>	STOP mode (or RESET)	1.65	-	-	V

1. Minimum supply voltage without losing data stored in RAM (in Stop mode or under Reset) or in hardware registers (only in Stop mode).



### Flash memory and data EEPROM

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
V <sub>DD</sub>	Operating voltage Read / Write / Erase	-	1.65	-	3.6	V
t <sub>prog</sub>	Programming time for word or half-page	Erasing	-	3.28	3.94	
		Programming	-	3.28	3.94	ms
I <sub>DD</sub>	Average current during the whole programming / erase operation	T <sub>A</sub> = 25 °C, V <sub>DD</sub> = 3.6 V	-	500	700	μA
	Maximum current (peak) during the whole programming / erase operation		-	1.5	2.5	mA

Table 43. Flash memory and data EEPROM characteristics

1. Guaranteed by design, not tested in production.

0h.al		0	Value	11
Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Unit
N <sub>CYC</sub> <sup>(2)</sup>	Cycling (erase / write) Program memory	T <sub>A</sub> = -40°C to 105 °C	10	kcycles
	Cycling (erase / write) EEPROM data memory		100	
	Cycling (erase / write) Program memory	T <sub>A</sub> = -40°C to 125 °C	0.2	
	Cycling (erase / write) EEPROM data memory		2	
	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 85 °C	-T <sub>RET</sub> = +85 °C	30	years
t <sub>RET</sub> <sup>(2)</sup>	Data retention (EEPROM data memory) after 100 kcycles at $T_A = 85 \text{ °C}$	TRET - 105 C	30	
	Data retention (program memory) after 10 kcycles at T <sub>A</sub> = 105 °C	T <sub>RFT</sub> = +105 °C	10	
	Data retention (EEPROM data memory) after 100 kcycles at $T_A = 105$ °C	TRET - +105 C		
	Data retention (program memory) after 200 cycles at T <sub>A</sub> = 125 °C	-T <sub>RET</sub> = +125 °C		
	Data retention (EEPROM data memory) after 2 kcycles at $T_A$ = 125 °C	RET - TIZS C		

### Table 44. Flash memory and data EEPROM endurance and retention

1. Guaranteed by characterization results, not tested in production.

2. Characterization is done according to JEDEC JESD22-A117.



### 6.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 45*. They are based on the EMS levels and classes defined in application note AN1709.

Symbol	Parameter	Conditions	Level/ Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP32, T <sub>A</sub> = +25 °C, f <sub>HCLK</sub> = 32 MHz conforms to IEC 61000-4-2	3B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on $V_{DD}$ and $V_{SS}$ pins to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP32, T <sub>A</sub> = +25 °C, f <sub>HCLK</sub> = 32 MHz conforms to IEC 61000-4-4	4A

### Table 45. EMS characteristics

### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

### Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical data corruption (control registers...)

### Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the oscillator pins for 1 second.



To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

#### **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. frequency range (32 MHz voltage Range 1)	Unit
S <sub>EMI</sub> Peak leve		k level $V_{DD} = 3.3 V$ , $T_A = 25 °C$ , LQFP32 package compliant with IEC 61967-2	0.1 to 30 MHz	-22	
	Peak level		30 to 130 MHz	-7	dBµV
			130 MHz to 1GHz	-12	
			SAE EMI Level	1	-

Table 46. EMI characteristics

#### 6.3.11 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts  $\times$  (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 47.	ESD	absolute	maximum	ratings
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Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	$T_A = +25 \text{ °C},$ conforming to ANSI/JEDEC JS-001	2	2000	v
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	$T_A = +25 \text{ °C},$ conforming to ANSI/ESD STM5.3.1.	C4	500	v

1. Guaranteed by characterization results, not tested in production.

#### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latch-up standard.



Symbol	Parameter	Conditions	Class			
LU	Static latch-up class	$T_A = +125$ °C conforming to JESD78A	II level A			

#### Table 48. Electrical sensitivities

#### 6.3.12 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DD}$  (for standard pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of  $-5 \mu$ A/+0  $\mu$ A range), or other functional failure (for example reset occurrence oscillator frequency deviation).

The test results are given in the Table 49.

		Functional s		
Symbol	Description	Negative injection	Positive injection	Unit
	Injected current on BOOT0	-0	NA <sup>(1)</sup>	
I <sub>INJ</sub>	Injected current on all FT pins	-5 <sup>(2)</sup>	NA <sup>(1)</sup>	mA
	Injected current on any other pin	-5 <sup>(2)</sup>	+5	

#### Table 49. I/O current injection susceptibility

1. Current injection is not possible.

2. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.



#### 6.3.13 I/O port characteristics

#### General input/output characteristics

Unless otherwise specified, the parameters given in *Table 50* are derived from tests performed under the conditions summarized in *Table 18*. All I/Os are CMOS and TTL compliant.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL</sub> V <sub>IH</sub>	Input low level voltage	TC, FT, FTf, RST I/Os	-	-	0.3V <sub>DD</sub>	
		BOOT0 pin	-	-	0.14V <sub>DD</sub> <sup>(1)</sup>	
	Input high lovel voltage	All I/Os except BOOT0 pin	0.7 V <sub>DD</sub>	-	-	V
	Input high level voltage	BOOT0 pin	0.15 V <sub>DD</sub> +0.56 <sup>(1)</sup>	-	-	
	I/O Schmitt trigger voltage hysteresis	Standard I/Os	-	10% V <sub>DD</sub> <sup>(3)</sup>	-	
V <sub>hys</sub>	(2)	BOOT0 pin	-	0.01	-	
l <sub>ikg</sub>		$\label{eq:VSS} \begin{array}{l} V_{SS} \leq V_{IN} \leq V_{DD} \\ \mbox{All I/Os except} \\ \mbox{BOOT0 and FTf} \\ \mbox{I/Os} \end{array}$	-	-	±50	nA
	Input leakage current <sup>(4)</sup>	BOOT0 <sup>(5)</sup> V <sub>IN</sub> = V <sub>DD</sub>	-	+2	-	
		BOOT0 V <sub>IN</sub> = V <sub>SS</sub>	-	0	-	μA
		$V_{DD} \le V_{IN} \le 5 V$ FT I/Os	-	-	200	nA
		V <sub>DD</sub> ≤ V <sub>IN</sub> ≤ 5 V FTf I/Os	-	-	500	
		$V_{DD} \le V_{IN} \le 5 V$ BOOT0	-	-	10	μA
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(6)</sup>	$V_{IN} = V_{SS}$	25	45	65	kΩ
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(6)</sup>	$V_{IN} = V_{DD}$	25	45	65	kΩ
C <sub>IO</sub>	I/O pin capacitance	-	-	5	-	pF

Table 50. I/O Static Characteristics	Table 50.	I/O	static	characteristics
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1. Guaranteed by characterization, not tested in production

2. Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results, not tested in production.

3. With a minimum of 200 mV. Guaranteed by characterization results, not tested in production.

4. The max. value may be exceeded if negative current is injected on adjacent pins.

5. BOOT0/PB9 pin limitation: typical input leakage current = 2  $\mu$ A and input frequency limited to 10 kHz (1.65 V < V<sub>DD</sub> < 2.7 V) and 5 MHz (2.7 V < V<sub>DD</sub> < 3.6 V).

 Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This MOS/NMOS contribution to the series resistance is minimum (~10% order).



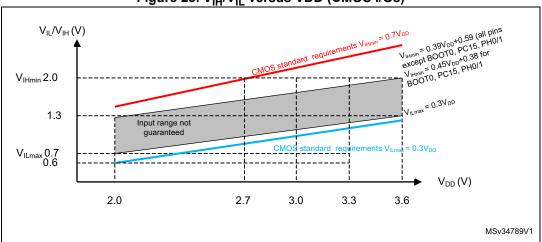
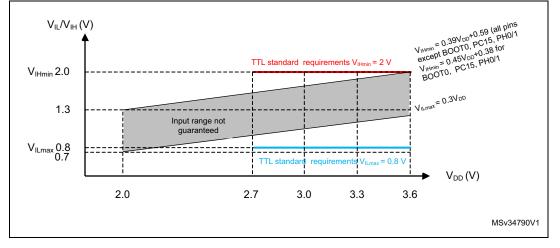


Figure 23. V<sub>IH</sub>/V<sub>IL</sub> versus VDD (CMOS I/Os)





#### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 15$  mA with the non-standard V<sub>OL</sub>/V<sub>OH</sub> specifications given in *Table 51*.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2:

- The sum of the currents sourced by all the I/Os on V<sub>DD</sub>, plus the maximum Run consumption of the MCU sourced on V<sub>DD</sub>, cannot exceed the absolute maximum rating I<sub>VDD(Σ)</sub> (see *Table 16*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub> plus the maximum Run consumption of the MCU sunk on V<sub>SS</sub> cannot exceed the absolute maximum rating I<sub>VSS(Σ)</sub> (see *Table 16*).



#### **Output voltage levels**

Unless otherwise specified, the parameters given in *Table 51* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18*. All I/Os are CMOS and TTL compliant.

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup> , I <sub>IO</sub> = +8 mA	-	0.4	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	$1_{O} = +0.11$ A 2.7 V $\leq V_{DD} \leq 3.6$ V	V <sub>DD</sub> -0.4	-	
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin	$\begin{array}{c} {\sf TTL \ port}^{(2)}, \\ {\sf I}_{IO} = + \ 8 \ mA \\ {\sf 2.7 \ V} \leq {\sf V}_{DD} \leq {\sf 3.6 \ V} \end{array}$	-	0.4	
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	$\begin{array}{c} \text{TTL port}^{(2)},\\ \text{I}_{\text{IO}} \texttt{=} -6 \text{ mA}\\ \texttt{2.7 V} \le \text{V}_{\text{DD}} \le \texttt{3.6 V} \end{array}$	2.4	-	
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	$I_{IO} = +15 \text{ mA}$ $2.7 \text{ V} \le V_{DD} \le 3.6 \text{ V}$	-	1.3	V
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	$I_{IO} = -15 \text{ mA}$ $2.7 \text{ V} \le V_{DD} \le 3.6 \text{ V}$	V <sub>DD</sub> -1.3	-	
V <sub>OL</sub> <sup>(1)(4)</sup>	Output low level voltage for an I/O pin	$I_{IO}$ = +4 mA 1.65 V $\leq$ V <sub>DD</sub> < 3.6 V	-	0.45	
V <sub>OH</sub> <sup>(3)(4)</sup>	Output high level voltage for an I/O pin	$I_{IO} = -4 \text{ mA}$ $1.65 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$	V <sub>DD</sub> -0.45	-	
V <sub>OLFM+</sub> <sup>(1)(4)</sup>	Output low level voltage for an FTf	$\begin{array}{c} \text{I}_{\text{IO}} = 20 \text{ mA} \\ 2.7 \text{ V} \leq \text{V}_{\text{DD}} \leq 3.6 \text{ V} \end{array}$	-	0.4	
* OLFM+	I/O pin in Fm+ mode	$\begin{array}{c} \text{I}_{\text{IO}} = 10 \text{ mA} \\ 1.65 \text{ V} \leq \text{V}_{DD} \leq 3.6 \text{ V} \end{array}$	-	0.4	

 The I<sub>IO</sub> current sunk by the device must always respect the absolute maximum rating specified in *Table 16*. The sum of the currents sunk by all the I/Os (I/O ports and control pins) must always be respected and must not exceed ΣI<sub>IO(PIN)</sub>.

2. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

3. The I<sub>IO</sub> current sourced by the device must always respect the absolute maximum rating specified in Table 16. The sum of the currents sourced by all the I/Os (I/O ports and control pins) must always be respected and must not exceed  $\Sigma I_{IO(PIN)}$ .

4. Guaranteed by characterization results, not tested in production.



#### Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 25* and *Table 52*, respectively.

Unless otherwise specified, the parameters given in *Table 52* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18*.

OSPEEDRx [1:0] bit value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Max <sup>(3)</sup>	Unit	
	f	Maximum frequency <sup>(4)</sup>	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	400	kHz	
00	f <sub>max(IO)out</sub>		$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	100	KI IZ	
00	t <sub>f(IO)out</sub>	Output rise and fall time	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	125	ns	
	t <sub>r(IO)out</sub>		$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	320	115	
	f	Maximum frequency <sup>(4)</sup>	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	2	MHz	
01	<sup>T</sup> max(IO)out	f <sub>max(IO)out</sub>		$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	0.6	
01	t <sub>f(IO)</sub> out t <sub>r(IO)</sub> out	<sup>ut</sup> Output rise and fall time	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	30	ns	
			$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	65	115	
	F	ut Maximum frequency <sup>(4)</sup>	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	10	MHz	
10	F <sub>max(IO)out</sub>	rmax(IO)out	Maximum frequency /	$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	2	
10	t <sub>f(IO)out</sub>	Output rise and fall time	$C_{L}$ = 50 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	13		
	t <sub>r(IO)out</sub> Output rise and fall	Output rise and fail time	$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	28	ns	
	-	Maximum fraguency (4)	$C_L = 30 \text{ pF}, V_{DD} = 2.7 \text{ V to } 3.6 \text{ V}$	-	35	MHz	
11	F <sub>max(IO)out</sub> Maximum frequency <sup>(4)</sup>	$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	10			
11	t <sub>f(IO)out</sub>		$C_{L}$ = 30 pF, $V_{DD}$ = 2.7 V to 3.6 V	-	6		
	t <sub>r(IO)out</sub>		$C_{L}$ = 50 pF, $V_{DD}$ = 1.65 V to 2.7 V	-	17	ns	
-	t <sub>EXTIpw</sub>	Pulse width of external signals detected by the EXTI controller	-	8	-	ns	

Table 52. I/O AC characteristics<sup>(1)(2)</sup>

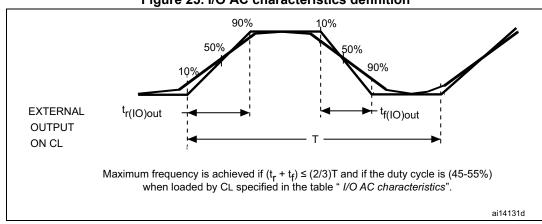
1. The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the line reference manual for a description of GPIO Port configuration register.

2. BOOT0/PB9 maximum input frequency is 10 kHz (1.65 V <  $V_{DD}$  < 2.7 V) and 5 MHz (2.7 V <  $V_{DD}$  < 3.6 V).

3. Guaranteed by design. Not tested in production.

4. The maximum frequency is defined in *Figure 25*.







#### 6.3.14 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor,  $R_{PU}$ , except when it is internally driven low (see *Table 53*).

Unless otherwise specified, the parameters given in *Table 53* are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in *Table 18*.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IL(NRST)</sub> <sup>(1)</sup>	NRST input low level voltage	-	-	-	$0.3V_{DD}$	
V <sub>IH(NRST)</sub> <sup>(1)</sup>	NRST input high level voltage	-	0.39V <sub>DD</sub> + 0.59	-	-	
V <sub>OL(NRST)</sub> <sup>(1)</sup>	NRST output low level voltage	I <sub>OL</sub> = 2 mA 2.7 V < V <sub>DD</sub> < 3.6 V	-	-	0.4	V
	inks i ouiput iow ievel voltage	I <sub>OL</sub> = 1.5 mA 1.65 V < V <sub>DD</sub> < 2.7 V	-	-	0.4	
V <sub>hys(NRST)</sub> <sup>(1)</sup>	NRST Schmitt trigger voltage hysteresis	-	-	10%V <sub>DD</sub> <sup>(2)</sup>	-	mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(3)</sup>	$V_{IN} = V_{SS}$	25	45	65	kΩ
V <sub>F(NRST)</sub> <sup>(1)</sup>	NRST input filtered pulse	-	-	-	50	ns
V <sub>NF(NRST)</sub> <sup>(1)</sup>	NRST input not filtered pulse	-	350	-	-	ns

Table 53. NRST pin characteristics

1. Guaranteed by design, not tested in production.

2. 200 mV minimum value

3. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is around 10%.



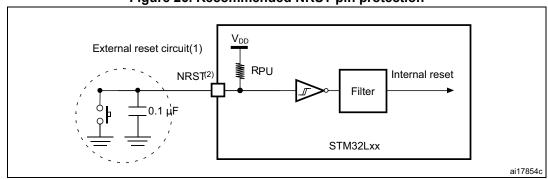


Figure 26. Recommended NRST pin protection

1. The reset network protects the device against parasitic resets.

- 2. The external capacitor must be placed as close as possible to the device.
- The user must ensure that the level on the NRST pin can go below the V<sub>IL(NRST)</sub> max level specified in Table 53. Otherwise the reset will not be taken into account by the device.

#### 6.3.15 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 54* are values derived from tests performed under ambient temperature,  $f_{PCLK}$  frequency and  $V_{DDA}$  supply voltage conditions summarized in *Table 18: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
V	Analog supply voltage for	Fast channel	1.65	-	3.6	V
V <sub>DDA</sub>	ADC ON	Standard channels	1.75 <sup>(1)</sup>	-	3.6	v
	Current consumption of the	1.14 Msps	-	200	-	
	ADC on V <sub>DDA</sub>	10 ksps	-	40	-	
IDDA (ADC)	Current consumption of the	1.14 Msps	-	70	-	μA
	ADC on V <sub>DD</sub> <sup>(2)</sup>	10 ksps	-	1	-	
f <sub>ADC</sub>		Voltage scaling Range 1	0.14	-	16	MHz
	ADC clock frequency	Voltage scaling Range 2	0.14	-	8	
		Voltage scaling Range 3	0.14	-	4	
f <sub>S</sub> <sup>(3)</sup>	Sampling rate	-	0.05	-	1.14	MHz
f <sub>TRIG</sub> <sup>(3)</sup>	External trigger frequency	f <sub>ADC</sub> = 16 MHz, 16-bit resolution	-	-	941	kHz
		-	-	-	17	1/f <sub>ADC</sub>
V <sub>AIN</sub>	Conversion voltage range	-	0	-	V <sub>DDA</sub>	V
R <sub>AIN</sub> <sup>(3)</sup>	External input impedance	See <i>Equation 1</i> and <i>Table 55</i> for details	-	-	50	kΩ
R <sub>ADC</sub> <sup>(3)(4)</sup>	Sampling switch resistance	-	-	-	1	kΩ

Table 54. ADC characteristics



Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
C <sub>ADC</sub> <sup>(3)</sup>	Internal sample and hold capacitor	-	-	-	8	pF
<b>↓</b> (3)	Calibration time	f <sub>ADC</sub> = 16 MHz		5.2		μs
t <sub>CAL</sub> <sup>(3)</sup>		-		83		1/f <sub>ADC</sub>
	W <sub>LATENCY</sub> ADC_DR register write latency	ADC clock = HSI16	1.5 ADC cycles + 2 f <sub>PCLK</sub> cycles	-	1.5 ADC cycles + 3 f <sub>PCLK</sub> cycles	-
W <sub>LATENCY</sub>		ADC clock = PCLK/2	-	4.5	-	f <sub>PCLK</sub> cycle
	ADC clock = PCLK/4	-	8.5	-	f <sub>PCLK</sub> cycle	
		$f_{ADC} = f_{PCLK}/2 = 16 \text{ MHz}$	0.266			μs
		$f_{ADC} = f_{PCLK}/2$	8.5			1/f <sub>PCLK</sub>
t <sub>latr</sub> (3)	Trigger conversion latency	$f_{ADC} = f_{PCLK}/4 = 8 \text{ MHz}$		0.516		μs
		f <sub>ADC</sub> = f <sub>PCLK</sub> /4		16.5		1/f <sub>PCLK</sub>
		f <sub>ADC</sub> = f <sub>HSI16</sub> = 16 MHz	0.252	-	0.260	μs
Jitter <sub>ADC</sub>	ADC jitter on trigger conversion	f <sub>ADC</sub> = f <sub>HSI16</sub>	-	1	-	1/f <sub>HSI16</sub>
ts <sup>(3)</sup>	Sampling time	f <sub>ADC</sub> = 16 MHz	0.093	-	10.03	μs
l <sub>S</sub> (-)		-	1.5	-	239.5	1/f <sub>ADC</sub>
t <sub>STAB</sub> <sup>(3)</sup>	Power-up time	-	0	0	1	μs
	Total conversion time	f <sub>ADC</sub> = 16 MHz	0.875		10.81	μs
t <sub>ConV</sub> <sup>(3)</sup>	(including sampling time)	-	14 to 173 (t <sub>S</sub> fo successive app			1/f <sub>ADC</sub>

#### Table 54. ADC characteristics (continued)

1. V<sub>DDA</sub> minimum value can be decreased in specific temperature conditions. Refer to *Table 55: RAIN max for fADC* = 16 *MHz*.

2. A current consumption proportional to the APB clock frequency has to be added (see *Table 32: Peripheral current consumption in run or Sleep mode*).

3. Guaranteed by design, not tested in production.

 Standard channels have an extra protection resistance which depends on supply voltage. Refer to Table 55: RAIN max for fADC = 16 MHz.

Equation 1:  $R_{AIN} \max_{T_{\alpha}} formula$ 

$$R_{AIN} < \frac{r_{S}}{f_{ADC} \times C_{ADC} \times \ln(2^{N+2})} - R_{ADC}$$

The simplified formula above (*Equation 1*) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).



		P may for			R <sub>AIN</sub> max	for stand	dard chan	inels (kΩ)	
T <sub>s</sub> (cycles)	t <sub>S</sub> (µs)	R <sub>AIN</sub> max for fast channels (kΩ)	V <sub>DD</sub> > 2.7 V	V <sub>DD</sub> > 2.4 V	V <sub>DD</sub> > 2.0 V	V <sub>DD</sub> > 1.8 V	V <sub>DD</sub> > 1.75 V	V <sub>DD</sub> > 1.65 V and T <sub>A</sub> > -10 °C	V <sub>DD</sub> > 1.65 V and T <sub>A</sub> > 25 °C
1.5	0.09	0.5	< 0.1	NA	NA	NA	NA	NA	NA
3.5	0.22	1	0.2	< 0.1	NA	NA	NA	NA	NA
7.5	0.47	2.5	1.7	1.5	< 0.1	NA	NA	NA	NA
12.5	0.78	4	3.2	3	1	NA	NA	NA	NA
19.5	1.22	6.5	5.7	5.5	3.5	NA	NA	NA	< 0.1
39.5	2.47	13	12.2	12	10	NA	NA	NA	5
79.5	4.97	27	26.2	26	24	< 0.1	NA	NA	19
160.5	10.03	50	49.2	49	47	32	< 0.1	< 0.1	42

### Table 55. $R_{AIN}$ max for $f_{ADC}$ = 16 MHz<sup>(1)</sup>

1. Guaranteed by design.

## Table 56. ADC accuracy $^{(1)(2)(3)(4)}$

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ET	Total unadjusted error		-	2	4	
EO	Offset error		-	1	2.5	
EG	Gain error		-	1	2	LSB
EL	Integral linearity error		-	1.5	2.5	
ED	Differential linearity error		-	1	1.5	
	Effective number of bits	1.65 V < V <sub>DDA</sub> < 3.6 V, range	10.2	11		
ENOB	Effective number of bits (16-bit mode oversampling with ratio =256) <sup>(5)</sup>	1/2/3, except for TSSOP14 package	11.3	12.1	-	bits
SINAD	Signal-to-noise distortion		62	67.8	-	
	Signal-to-noise ratio		63	68	-	
SNR	Signal-to-noise ratio (16-bit mode oversampling with ratio =256) <sup>(5)</sup>	·	70	76	-	dB
THD	Total harmonic distortion		-	-81	-68.5	



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
ET	Total unadjusted error		-	3	5	
EO	Offset error		-	2	2.5	
EG	Gain error		-	2	2.5	LSB
EL	Integral linearity error		-	1.5	2.5	
ED	Differential linearity error		-	1	1.7	
	Effective number of bits	1.65 V < V <sub>DDA</sub> < 3.6 V, range 1/2/3, TSSOP14 package	9.5	10.5	-	
ENOB	Effective number of bits (16-bit mode oversampling with ratio =256) <sup>(5)</sup>		10.7	11.6	-	bits
SINAD	Signal-to-noise distortion		59	65	-	
	Signal-to-noise ratio		59	65	-	
SNR	Signal-to-noise ratio (16-bit mode oversampling with ratio =256) <sup>(5)</sup>		66	73	-	dB
THD	Total harmonic distortion		-	-75	-63	

#### Table 56. ADC $accuracy^{(1)(2)(3)(4)}$

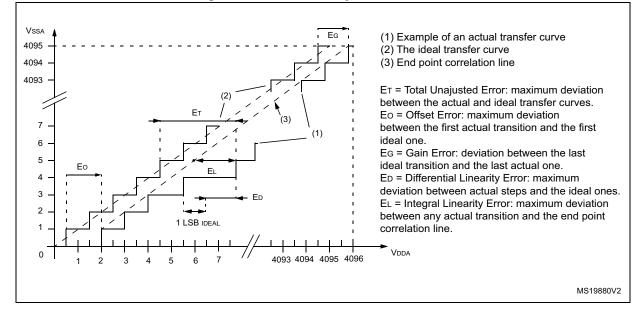
1. ADC DC accuracy values are measured after internal calibration.

 ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current. Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.12 does not affect the ADC

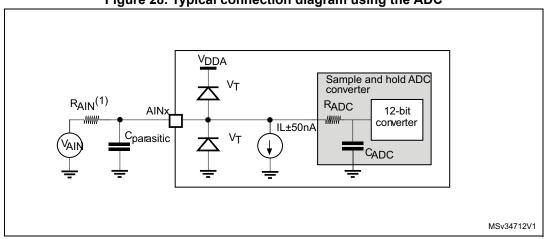
Any positive injection current within the limits specified for  $I_{INJ(PIN)}$  and  $\Sigma I_{INJ(PIN)}$  in Section 6.3.12 does not affect the ADC accuracy.

- 3. Better performance may be achieved in restricted  $V_{\text{DDA}}$ , frequency and temperature ranges.
- 4. In TSSOP14 package, where V<sub>DDA</sub> pin is shared with V<sub>DD</sub> pin, I/O toggling should be minimized to reach the values given in the above table. I/O toggling with loaded I/O pins can generate ripple on V<sub>DD</sub>/V<sub>DDA</sub> and degrade the ADC accuracy.
- 5. This number is obtained by the test board without additional noise, resulting in non-optimized value for oversampling mode.

#### Figure 27. ADC accuracy characteristics









- 1. Refer to Table 54: ADC characteristics for the values of RAIN, RADC and CADC.
- C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.

#### 6.3.16 Temperature sensor characteristics

#### Table 57. Temperature sensor calibration values

Calibration value name	Description	Memory address
TS_CAL2	TS ADC raw data acquired at temperature of 130 °C $\pm$ 5 °C, V <sub>DDA</sub> = 3 V $\pm$ 10 mV	0x1FF8 007E - 0x1FF8 007F

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with temperature	-	±1	<u>+2</u>	°C
Avg_Slope <sup>(1)</sup>	Average slope	1.48	1.61	1.75	mV/°C
V <sub>130</sub>	Voltage at 130°C ±5°C <sup>(2)</sup>	640	670	700	mV
I <sub>DDA(TEMP)</sub> <sup>(3)</sup>	Current consumption	-	3.4	6	μA
t <sub>START</sub> <sup>(3)</sup>	Startup time	-	-	10	
T <sub>S_temp</sub> <sup>(4)(3)</sup>	ADC sampling time when reading the temperature	10	-	-	μs

#### Table 58. Temperature sensor characteristics

1. Guaranteed by characterization results, not tested in production.

2. Measured at  $V_{DD}$  = 3 V ±10 mV. V30 ADC conversion result is stored in the TS\_CAL1 byte.

3. Guaranteed by design, not tested in production.

4. Shortest sampling time can be determined in the application by multiple iterations.



#### 6.3.17 Comparators

Table 59. Comparator 1 characteristics							
Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Тур	Max <sup>(1)</sup>	Unit	
$V_{DDA}$	Analog supply voltage	-	1.65		3.6	V	
R <sub>400K</sub>	R <sub>400K</sub> value	-	-	400	-	kΩ	
R <sub>10K</sub>	R <sub>10K</sub> value	-	-	10	-	K22	
V <sub>IN</sub>	Comparator 1 input voltage range	-	0.6	-	V <sub>DDA</sub>	V	
t <sub>START</sub>	Comparator startup time	-	-	7	10		
td	Propagation delay <sup>(2)</sup>	-	-	3	10	μs	
V <sub>offset</sub>	Comparator offset <sup>(3)</sup>	-	-	±3	±10	mV	
d <sub>Voffset</sub> /dt	Comparator offset variation in worst voltage stress conditions <sup>(3)</sup>	$V_{DDA} = 3.6 V$ $V_{IN+} = 0 V$ $V_{IN-} = V_{REFINT}$ $T_A = 25 ° C$	0	1.5	10	mV/1000 h	
I <sub>COMP1</sub>	Current consumption <sup>(4)</sup>	-	-	160	260	nA	

#### Table 59. Comparator 1 characteristics

1. Guaranteed by characterization, not tested in production.

2. The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

 In TSSOP14 package, where V<sub>DDA</sub> pin is shared with V<sub>DD</sub> pin, I/O toggling should be minimized to reach the values given in the above table. I/O toggling with loaded I/O pins can generate ripple on V<sub>DD</sub>/V<sub>DDA</sub> and degrade the comparator performance.

4. Comparator consumption only. Internal reference voltage not included.

#### Table 60. Comparator 2 characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
V <sub>DDA</sub>	Analog supply voltage	-	1.65	-	3.6	V
V <sub>IN</sub>	Comparator 2 input voltage range	-	0	-	$V_{DDA}$	V
+.	Comparator startup time	Fast mode	-	15	-         3.6           -         V <sub>DDA</sub> 15         20           20         25           1.8         3.5           2.5         6	
t <sub>START</sub>		Slow mode	-	20	25	
+	Propagation delay <sup>(2)</sup> in slow mode	$1.65 \text{ V} \leq \text{V}_{\text{DDA}} \leq 2.7 \text{ V}$	-	1.8	3.5	
t <sub>d slow</sub>		$2.7 \text{ V} \leq \text{V}_{\text{DDA}} \leq 3.6 \text{ V}$	-	2.5	2.5 6	μs
4	Propagation delay <sup>(2)</sup> in fast mode	$1.65 \text{ V} \leq \text{V}_{\text{DDA}} \leq 2.7 \text{ V}$	-	0.8	2	
t <sub>d fast</sub>	Propagation delay / in last mode	$2.7 \text{ V} \leq \text{V}_{\text{DDA}} \leq 3.6 \text{ V}$	-	1.2	4	
V <sub>offset</sub>	Comparator offset error <sup>(3)</sup>		-	±4	±20	mV
dThreshold/ dt	Threshold voltage temperature coefficient	$V_{DDA} = 3.3V$ $T_{A} = 0 \text{ to } 50 \circ C$ $V = V_{REFINT},$ $3/4 V_{REFINT},$ $1/2 V_{REFINT},$ $1/4 V_{REFINT}.$	-	15	30	ppm /°C



Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
	Current consumption <sup>(4)</sup>	Fast mode	-	3.5	5	
ICOMP2		,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,	2	μA		

#### Table 60. Comparator 2 characteristics (continued)

1. Guaranteed by characterization results, not tested in production.

4. Comparator consumption only. Internal reference voltage (necessary for comparator operation) is not included.

#### 6.3.18 Timer characteristics

#### **TIM timer characteristics**

The parameters given in the Table 61 are guaranteed by design.

Refer to Section 6.3.13: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Symbol	Parameter	Conditions	Min	Мах	Unit
+	Timer resolution time		1	-	t <sub>TIMxCLK</sub>
<sup>t</sup> res(TIM)		f <sub>TIMxCLK</sub> = 32 MHz	31.25	-	ns
f	Timer external clock		0 f <sub>TIMxCLK</sub> /2 MHz 0 16 16	f <sub>TIMxCLK</sub> /2	MHz
f <sub>EXT</sub>	frequency on CH1 to CH4	f <sub>TIMxCLK</sub> = 32 MHz	0	16	MHz
Res <sub>TIM</sub>	Timer resolution	-		16	bit
	16-bit counter clock	-	1	65536	t <sub>TIMxCLK</sub>
<sup>t</sup> COUNTER	period when internal clock is selected (timer's prescaler disabled)	f <sub>TIMxCLK</sub> = 32 MHz	1       -         MHz       31.25       -         0       f <sub>TIMxCLK</sub> /2         MHz       0       16         1       65536         MHz       0.0312       2048         -       65536 × 65536	μs	
4	Maximum possible count	_	-	65536 × 65536	t <sub>TIMxCLK</sub>
<sup>t</sup> MAX_COUNT		f <sub>TIMxCLK</sub> = 32 MHz	-	134.2	S

Table 61. TIMx<sup>(1)</sup> characteristics

1. TIMx is used as a general term to refer to the TIM2 and TIM21 timers.



<sup>2.</sup> The delay is characterized for 100 mV input step with 10 mV overdrive on the inverting input, the non-inverting input set to the reference.

In TSSOP14 package, where V<sub>DDA</sub> pin is shared with V<sub>DD</sub> pin, I/O toggling should be minimized to reach the values given in the above table. I/O toggling with loaded I/O pins can generate ripple on V<sub>DD</sub>/V<sub>DDA</sub> and degrade the comparator performance.

#### 6.3.19 Communications interfaces

#### I<sup>2</sup>C interface characteristics

The  $I^2C$  interface meets the timings requirements of the  $I^2C$ -bus specification and user manual rev. 03 for:

- Standard-mode (Sm) : with a bit rate up to 100 kbit/s
- Fast-mode (Fm) : with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+) : with a bit rate up to 1 Mbit/s.

The I<sup>2</sup>C timing requirements are guaranteed by design when the I<sup>2</sup>C peripheral is properly configured (refer to the reference manual for details) and when the I2CCLK frequency is greater than the minimum given in *Table 63*. The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and VDDIOx is disabled, but is still present. Only FTf I/O pins support Fm+ low level output current maximum requirement (refer to *Section 6.3.13: I/O port characteristics* for the I2C I/Os characteristics).

All I<sup>2</sup>C SDA and SCL I/Os embed an analog filter (see *Table 62* for the analog filter characteristics).

The analog spike filter is compliant with I<sup>2</sup>C timings requirements only for the following voltage ranges:

- Fast mode Plus: 2.7 V ≤V<sub>DD</sub> ≤3.6 V and voltage scaling Range 1
- Fast mode:
  - 2 V ≤V<sub>DD</sub> ≤3.6 V and voltage scaling Range 1 or Range 2.
  - V<sub>DD</sub> < 2 V, voltage scaling Range 1 or Range 2, C<sub>load</sub> < 200 pF.</li>

In other ranges, the analog filter should be disabled. The digital filter can be used instead.

Note: In Standard mode, no spike filter is required.

Symbol	Parameter	Conditions	Min	Max	Unit
	t <sub>AF</sub> Maximum pulse width of spikes that are suppressed by the analog filter	Range 1		260 <sup>(3)</sup>	
t <sub>AF</sub>		Range 2	50 <sup>(2)</sup>	-	ns
		Range 3	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$		

Table 62. I2C analog filter characteristics<sup>(1)</sup>

1. Guaranteed by characterization results.

2. Spikes with widths below t<sub>AF(min)</sub> are filtered.

3. Spikes with widths above t<sub>AF(max)</sub> are not filtered



#### **Electrical characteristics**

Symbol	Parameter	Co	Min	Unit	
		Standard-mode		2	
	I2C clock frequency	Fast-mode		8	
f <sub>I2CCLK</sub>		East mode Plus	Analog filter ON, DNF = 0	18	MHz
		Fast-mode 8 Analog filter ON, 18			

Table 63. I2C frequency in all I2C modes

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#### **SPI characteristics**

Unless otherwise specified, the parameters given in the following tables are derived from tests performed under ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in *Table 18*.

Refer to *Section 6.3.12: I/O current injection characteristics* for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode			16	
		Slave mode receiver		-	16	
f <sub>SCK</sub> 1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode Transmitter 1.71 <v<sub>DD&lt;3.6V</v<sub>	-	-	12 <sup>(2)</sup>	MHz
		Slave mode Transmitter 2.7 <v<sub>DD&lt;3.6V</v<sub>	-	-	16 <sup>(2)</sup>	
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2Tpclk	-	-	
t <sub>w(SCKH)</sub> t <sub>w(SCKL)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2	
t <sub>su(MI)</sub>	Data input setup time	Master mode	3	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	3	-	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	3.5	-	-	
t <sub>h(SI)</sub>		Slave mode	0	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	15	-	36	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	10	-	30	
+		Slave mode 1.71 <v<sub>DD&lt;3.6V</v<sub>	-	14	35	
t <sub>v(SO)</sub>	Data output valid time	Data output valid time Slave mode 2.7 <v<sub>DD&lt;3.6V</v<sub>	-	14	20	
t <sub>v(MO)</sub>		Master mode	-	4	6	
t <sub>h(SO)</sub>	Data output hold time	Slave mode	10	-	-	
t <sub>h(MO)</sub>		Master mode	3	-	-	

Table 64, SPI characteris	tics in voltage Range 1 <sup>(1)</sup>
Table 64. SPI characteris	lics in voltage Range 1 V

1. Guaranteed by characterization results, not tested in production.

2. The maximum SPI clock frequency in slave transmitter mode is determined by the sum of  $t_{y(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty<sub>(SCK)</sub> = 50%.



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
		Master mode			8	
f <sub>SCK</sub> 1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode Transmitter 1.65 <v<sub>DD&lt;3.6V</v<sub>	-	-	8	MHz
		Slave mode Transmitter 2.7 <v<sub>DD&lt;3.6V</v<sub>			8(2)	
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2Tpclk	-	-	
t <sub>w(SCKH)</sub> t <sub>w(SCKL)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2	
t <sub>su(MI)</sub>	Dete innut estur time	Master mode	3	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	3	-	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	6	-	-	
t <sub>h(SI)</sub>		Slave mode	2	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	18	-	52	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	12	-	42	
	Data output valid time	Slave mode	-	16	33	
t <sub>v(SO)</sub>		Master mode	-	4	6	
t <sub>v(MO)</sub>	Data output hold time	Slave mode	11	-	-	1
t <sub>h(SO)</sub>		Master mode	3	-	-	

1. Guaranteed by characterization results, not tested in production.

2. The maximum SPI clock frequency in slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty<sub>(SCK)</sub> = 50%.

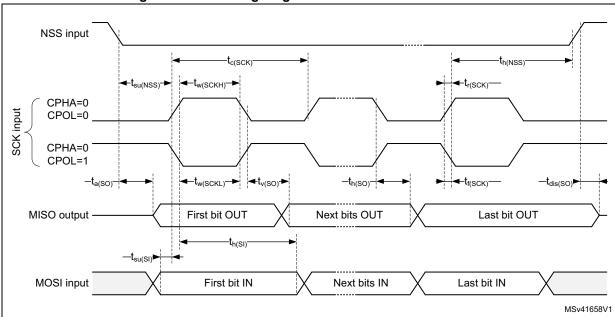


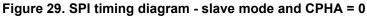
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>SCK</sub>	SDI clock frequency	Master mode			2	MHz
1/t <sub>c(SCK)</sub>	SPI clock frequency	Slave mode	-	-	2 <sup>(2)</sup>	
Duty <sub>(SCK)</sub>	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
t <sub>su(NSS)</sub>	NSS setup time	Slave mode, SPI presc = 2	4Tpclk	-	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode, SPI presc = 2	2Tpclk	-	-	
t <sub>w(SCKH)</sub> t <sub>w(SCKL)</sub>	SCK high and low time	Master mode	Tpclk-2	Tpclk	Tpclk+2	
t <sub>su(MI)</sub>	Data input setup time	Master mode	3	-	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	3	-	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	16	-	-	
t <sub>h(SI)</sub>	Data input hold time	Slave mode	14	-	-	ns
t <sub>a(SO</sub>	Data output access time	Slave mode	30	-	70	
t <sub>dis(SO)</sub>	Data output disable time	Slave mode	40	-	80	
t (ap)	Data output valid time	Slave mode	-	26.5	47	
t <sub>v(SO)</sub>		Master mode	-	4	6	
t <sub>v(MO)</sub>	Data autout hald time	Slave mode	20	-	-	
t <sub>h(SO)</sub>	Data output hold time	Master mode	3	-	-	1

1. Guaranteed by characterization results, not tested in production.

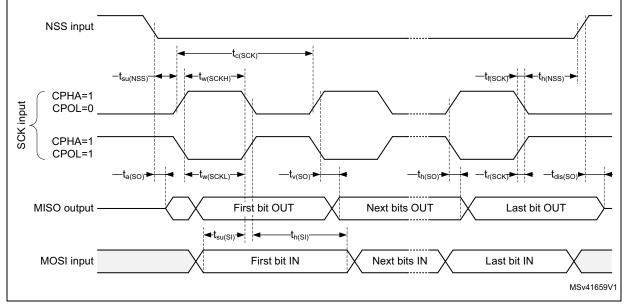
2. The maximum SPI clock frequency in slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty<sub>(SCK)</sub> = 50%.











1. Measurement points are done at CMOS levels:  $0.3V_{\text{DD}}$  and  $0.7V_{\text{DD}}$ 



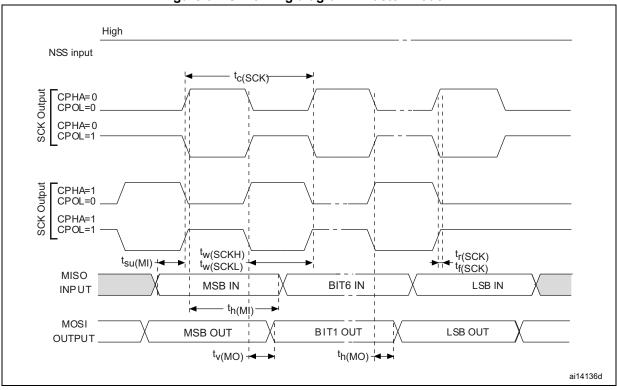


Figure 31. SPI timing diagram - master mode<sup>(1)</sup>

1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .



### 7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status *are available at http://www.st.com.* ECOPACK<sup>®</sup> is an ST trademark.

### 7.1 LQFP32 package information

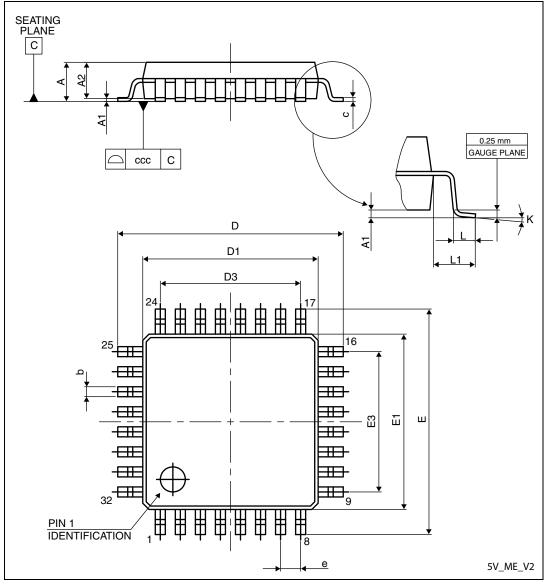


Figure 32. LQFP32 - 32-pin, 7 x 7 mm, 32-pin low-profile quad flat package outline

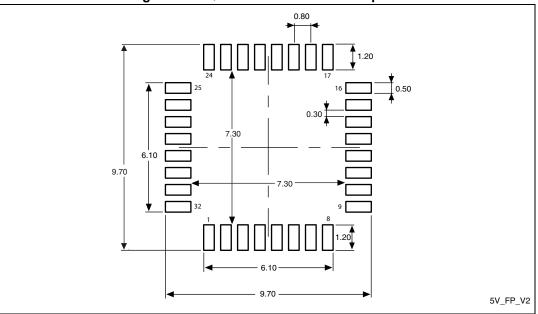
1. Drawing is not to scale.



mechanical data							
Cumhal		millimeters			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Мах	
А	-	-	1.600	-	-	0.0630	
A1	0.050	-	0.150	0.0020	-	0.0059	
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571	
b	0.300	0.370	0.450	0.0118	0.0146	0.0177	
С	0.090	-	0.200	0.0035	-	0.0079	
D	8.800	9.000	9.200	0.3465	0.3543	0.3622	
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835	
D3	-	5.600	-	-	0.2205	-	
Е	8.800	9.000	9.200	0.3465	0.3543	0.3622	
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835	
E3	-	5.600	-	-	0.2205	-	
е	-	0.800	-	-	0.0315	-	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295	
L1	-	1.000	-	-	0.0394	-	
CCC	-	-	0.100	-	-	0.0039	
k	0°	3.5°	7°	0°	3.5°	7°	

Table 67. LQFP32 - 32-pin, 7 x 7 mm, 32-pin low-profile quad flat package
mechanical data

1. Values in inches are converted from mm and rounded to 4 decimal digits.



#### Figure 33. LQFP32 recommended footprint

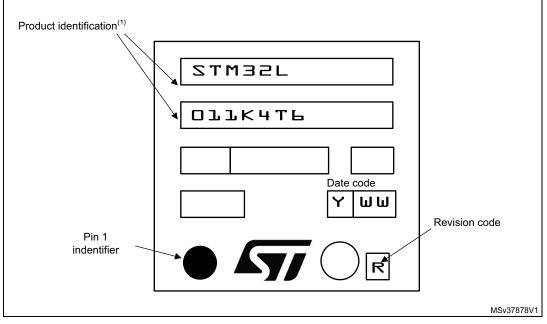
1. Dimensions are expressed in millimeters.

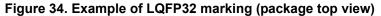


#### LQFP32 device marking

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.

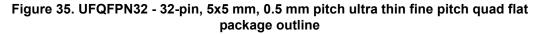


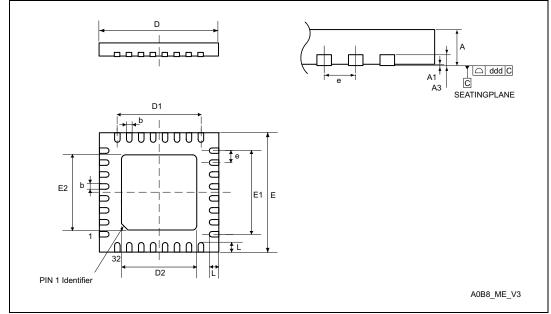


 Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.



## 7.2 UFQFPN32 package information





1. Drawing is not to scale.

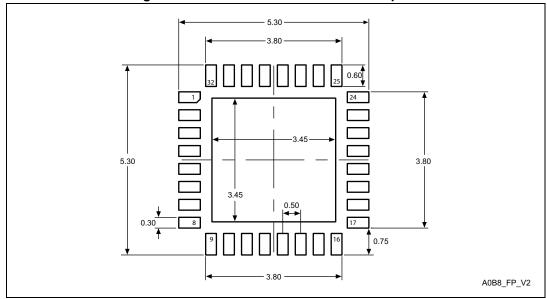
2. There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this backside pad to PCB ground.



package mechanical data								
Cumhal		millimeters			inches <sup>(1)</sup>			
Symbol	Min	Тур	Max	Min	Тур	Max		
А	0.500	0.550	0.600	0.0197	0.0217	0.0236		
A1	-	-	0.050	-	-	0.0020		
A3	-	0.152	-	-	0.0060	-		
b	0.180	0.230	0.280	0.0071	0.0091	0.0110		
D	4.900	5.000	5.100	0.1929	0.1969	0.2008		
D1	3.400	3.500	3.600	0.1339	0.1378	0.1417		
D2	3.400	3.500	3.600	0.1339	0.1378	0.1417		
Е	4.900	5.000	5.100	0.1929	0.1969	0.2008		
E1	3.400	3.500	3.600	0.1339	0.1378	0.1417		
E2	3.400	3.500	3.600	0.1339	0.1378	0.1417		
е	-	0.500	-	-	0.0197	-		
L	0.300	0.400	0.500	0.0118	0.0157	0.0197		
ddd	-	-	0.080	-	-	0.0031		

# Table 68. UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flatpackage mechanical data

1. Values in inches are converted from mm and rounded to 4 decimal digits.



#### Figure 36. UFQFPN32 recommended footprint

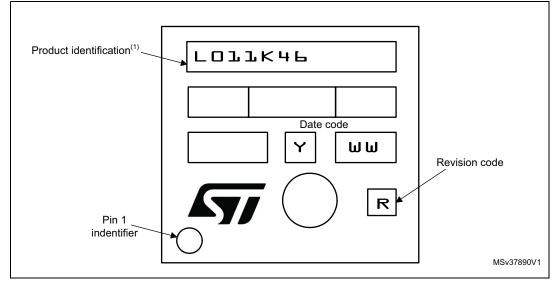
1. Dimensions are expressed in millimeters.



#### **Device marking**

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.





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### 7.3 WLCSP25 package information

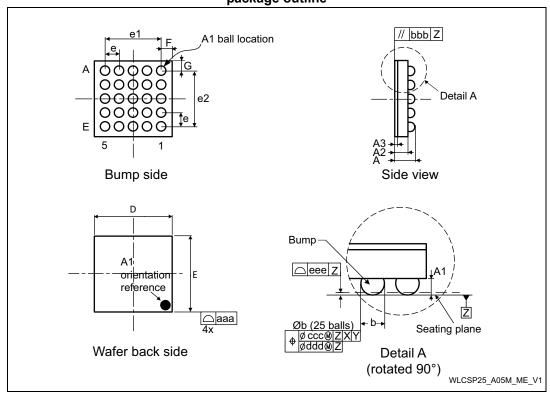


Figure 38. WLCSP25 - 25-ball, 2.133 x 2.070 mm, 0.4 mm pitch wafer level chip scale package outline

1. Drawing is not to scale.

Table 69. WLCSP25 - 25-ball, 2.133 x 2.070 mm, 0.4 mm pitch wafer level chip scale	
package mechanical data	

Gumbal		millimeters			inches <sup>(1)</sup>	
Symbol	Min	Тур	Мах	Min	Тур	Мах
А	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3 <sup>(2)</sup>	-	0.025	-	-	0.0010	-
b <sup>(3)</sup>	0.220	0.250	0.280	0.0087	0.0098	0.0110
D	2.098	2.133	2.168	0.0826	0.0840	0.0854
E	2.035	2.070	2.105	0.0801	0.0815	0.0829
е	-	0.400	-	-	0.0157	-
e1	-	1.600	-	-	0.0630	-
e2	-	1.600	-	-	0.0630	-
F	-	0.2665	-	-	0.0105	-



## Table 69. WLCSP25 - 25-ball, 2.133 x 2.070 mm, 0.4 mm pitch wafer level chip scale package mechanical data (continued)

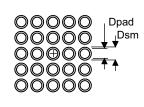
Symbol		millimeters		inches <sup>(1)</sup>		
	Min	Тур	Мах	Min	Тур	Мах
G	-	0.235	-	-	0.0093	-
aaa	-	-	0.100	-	-	0.0039
bbb	-	-	0.100	-	-	0.0039
CCC	-	-	0.100	-	-	0.0039
ddd	-	-	0.050	-	-	0.0020
eee	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.

2. Back side coating.

3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

## Figure 39. WLCSP25 - 25-ball, 2.133 x 2.070 mm, 0.4 mm pitch wafer level chip scale package recommended footprint



WLCSP25\_A05M\_FP\_V1

#### Table 70. WLCSP25 recommended PCB design rules (0.4 mm pitch)

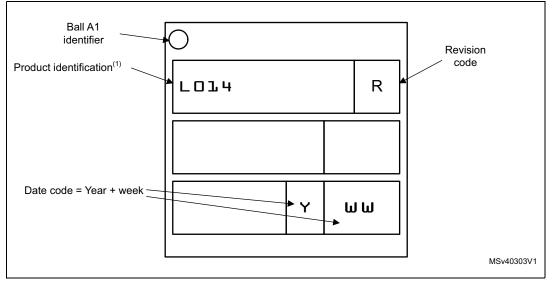
Dimension	Recommended values
Pitch	0.4 mm
Dpad	0.225 mm
Dsm	0.290 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm



#### **Device marking**

The following figure gives an example of topside marking versus ball A1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.





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### 7.4 UFQFPN28 4 x 4 mm package information

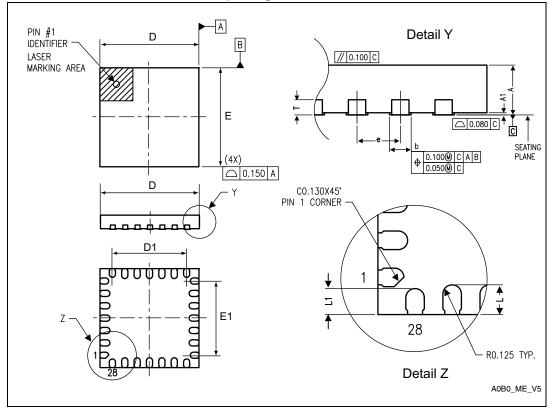


Figure 41. UFQFPN28 - 28-lead, 4x4 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package outline

1. Drawing is not to scale.

Table 71. UFQFPN28 - 28-lead, 4x4 mm, 0.5 mm pitch, ultra thin fine pitch quad flat
package mechanical data <sup>(1)</sup>

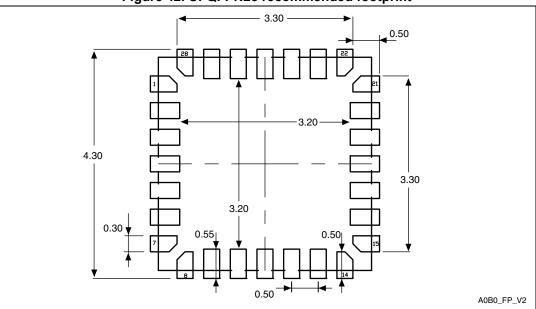
Symbol	millimeters			inches			
Symbol	Min	Тур	Max	Min	Тур	Мах	
A	0.500	0.550	0.600	0.0197	0.0217	0.0236	
A1	-	0.000	0.050	-	0.0000	0.0020	
D	3.900	4.000	4.100	0.1535	0.1575	0.1614	
D1	2.900	3.000	3.100	0.1142	0.1181	0.1220	
E	3.900	4.000	4.100	0.1535	0.1575	0.1614	
E1	2.900	3.000	3.100	0.1142	0.1181	0.1220	
L	0.300	0.400	0.500	0.0118	0.0157	0.0197	
L1	0.250	0.350	0.450	0.0098	0.0138	0.0177	
Т	-	0.152	-	-	0.0060	-	



# Table 71. UFQFPN28 - 28-lead, 4x4 mm, 0.5 mm pitch, ultra thin fine pitch quad flatpackage mechanical data<sup>(1)</sup>

Symbol		millimeters		inches		
Symbol	Min	Min Typ Max		Min Typ		Max
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.



#### Figure 42. UFQFPN28 recommended footprint

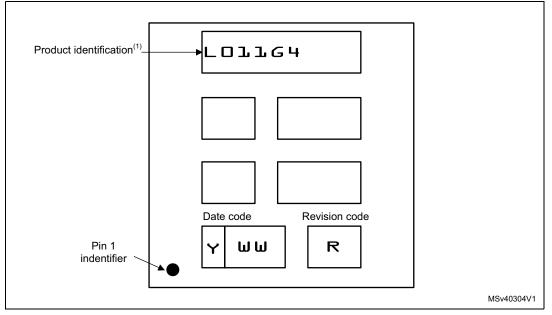
1. Dimensions are expressed in millimeters.

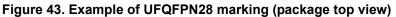


#### **UFQFPN28** device marking

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.

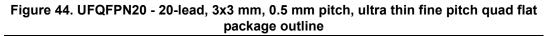


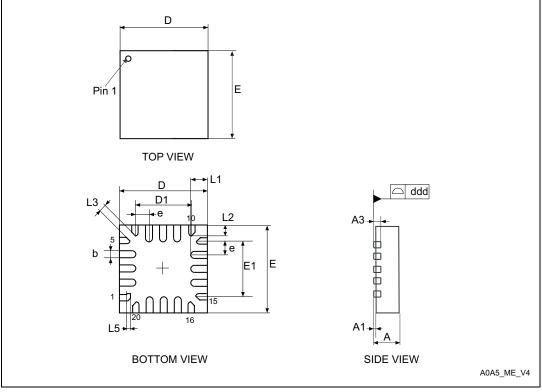


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## 7.5 UFQFPN20 package information





1. Drawing is not to scale.

Table 72. UFQFPN20 - 20-lead, 3x3 mm, 0.5 mm pitch, ultra thin fine pitch quad flat
package mechanical data

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Мах
А	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
A3	-	0.152	-	-	0.060	-
D	2.900	3.000	3.100	0.1142	0.1181	0.1220
D1	-	2.000	-	-	0.0790	-
E	2.900	3.000	3.100	0.1142	0.1181	0.1220
E1	-	2.000	-	-	0.0790	-
L1	0.500	0.550	0.600	0.0197	0.0217	0.0236
L2	0.300	0.350	0.400	0.0118	0.0138	0.0157
L3	-	0.200	-		0.0079	
L5	-	0.150	-		0.0059	

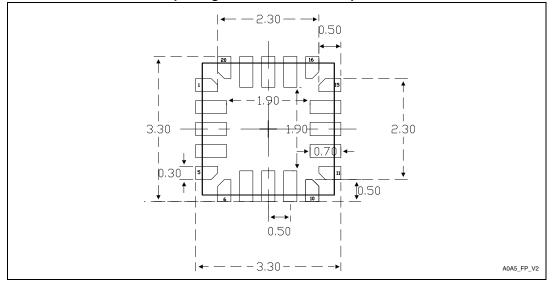


# Table 72. UFQFPN20 - 20-lead, 3x3 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package mechanical data (continued)

Symbol		millimeters		inches <sup>(1)</sup>		
Symbol	Min	Тур	Мах	Min	Тур	Мах
b	0.180	0.250	0.300	0.0071	0.0098	0.0118
е	-	0.500	-	-	0.0197	-
ddd	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.

## Figure 45. UFQFPN20 - 20-lead, 3x3 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package recommended footprint



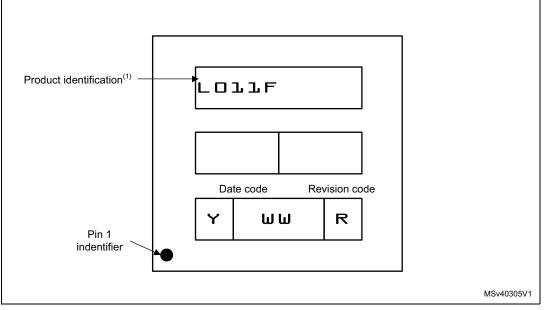
1. Dimensions are expressed in millimeters.

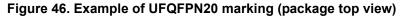


#### **UFQFPN20** device marking

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.

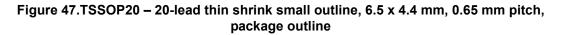


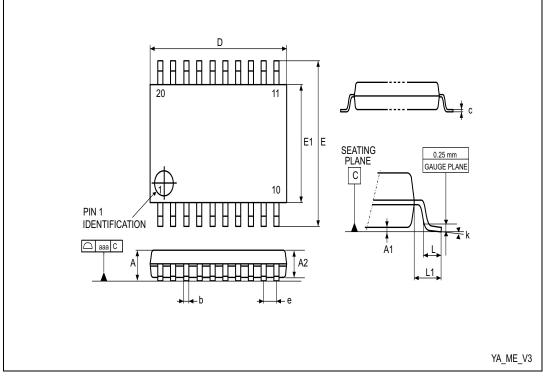


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## 7.6 TSSOP20 package information





1. Drawing is not to scale.

Table 73. TSSOP20 – 20-lead thin shrink small outline, 6.5 x 4.4 mm, 0.65 mm pitch,
package mechanical data

Cumhal	millimeters			inches <sup>(1)</sup>		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
А	-	-	1.200	-	-	0.0472
A1	0.050	-	0.150	0.0020	-	0.0059
A2	0.800	1.000	1.050	0.0315	0.0394	0.0413
b	0.190	-	0.300	0.0075	-	0.0118
С	0.090	-	0.200	0.0035	-	0.0079
D <sup>(2)</sup>	6.400	6.500	6.600	0.2520	0.2559	0.2598
Е	6.200	6.400	6.600	0.2441	0.2520	0.2598
E1 <sup>(3)</sup>	4.300	4.400	4.500	0.1693	0.1732	0.1772
е	-	0.650	-	-	0.0256	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-



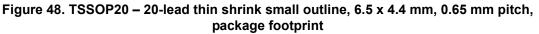
## Table 73. TSSOP20 – 20-lead thin shrink small outline, 6.5 x 4.4 mm, 0.65 mm pitch, package mechanical data (continued)

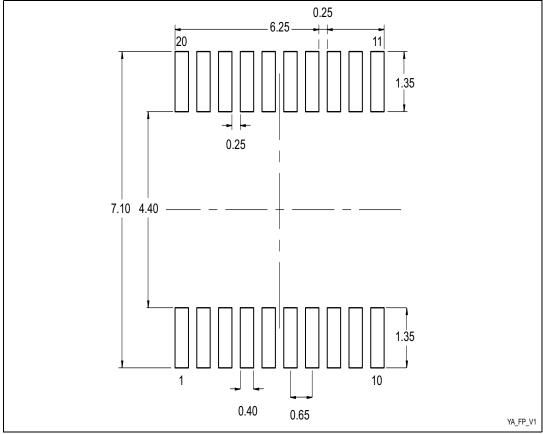
Symbol	millimeters			inches <sup>(1)</sup>			
Symbol	Min. Typ.		Max.	Min. Typ.		Max.	
k	0°	-	8°	0°	-	8°	
ааа	-	-	0.100	-	-	0.0039	

1. Values in inches are converted from mm and rounded to four decimal digits.

2. Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs shall not exceed 0.15mm per side.

3. Dimension "E1" does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.





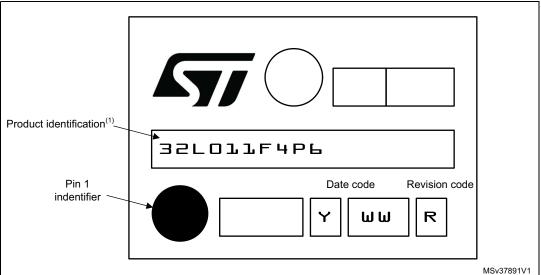
1. Dimensions are expressed in millimeters.



#### **Device marking**

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.

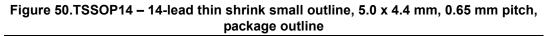


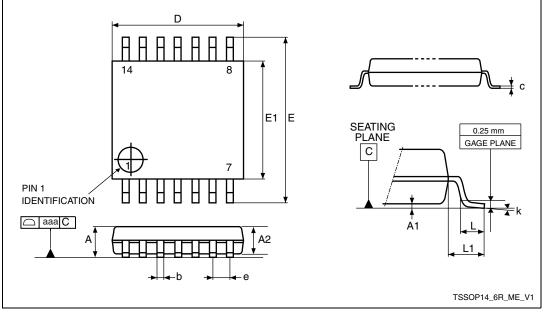


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### 7.7 TSSOP14 package information





1. Drawing is not to scale.

## Table 74. TSSOP14 – 14-lead thin shrink small outline, 5 x 4.4 mm, 0.65 mm pitch,package mechanical data

Cumb al		millimeters		inches <sup>(1)</sup>			
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.	
А	-	-	1.200	-	-	0.0472	
A1	0.050	-	0.150	0.0020	-	0.0059	
A2	0.800	1.000	1.050	0.0315	0.0394	0.0413	
b	0.190	-	0.300	0.0075	-	0.0118	
С	0.090	-	0.200	0.0035	-	0.0079	
D	4.900	5.000	5.100	0.1929	0.1969	0.2008	
E	6.200	6.400	6.600	0.2441	0.2520	0.2598	
E1	4.300	4.400	4.500	0.1693	0.1732	0.1772	
е	-	0.650	-	-	0.0256	-	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295	
L1	-	1.000	-	-	0.0394	-	
k	0°	-	8°	0°	-	8°	
ааа	-	-	0.100	-	-	0.0039	

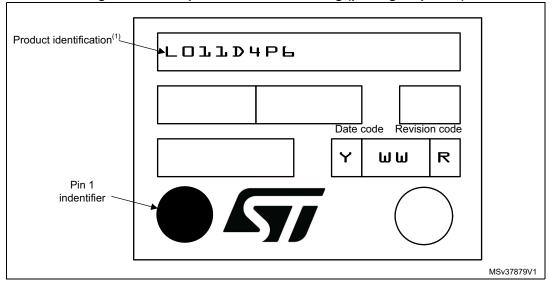
1. Values in inches are converted from mm and rounded to four decimal digits.

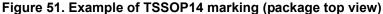


#### **TSSOP14** device marking

The following figure gives an example of topside marking versus pin 1 position identifier location.

Other optional marking or inset/upset marks, which depends assembly location, are not indicated below.





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### 7.8 Thermal characteristics

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

 $T_J \max = T_A \max + (P_D \max \times \Theta_{JA})$ 

Where:

- T<sub>A</sub> max is the maximum ambient temperature in °C,
- O<sub>JA</sub> is the package junction-to-ambient thermal resistance, in °C/W,
- P<sub>D</sub> max is the sum of P<sub>INT</sub> max and P<sub>I/O</sub> max (P<sub>D</sub> max = P<sub>INT</sub> max + P<sub>I/O</sub>max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

P<sub>I/O</sub> max represents the maximum power dissipation on output pins where:

 $\mathsf{P}_{\mathsf{I}/\mathsf{O}} \max = \Sigma \; (\mathsf{V}_{\mathsf{OL}} \times \mathsf{I}_{\mathsf{OL}}) + \Sigma ((\mathsf{V}_{\mathsf{DD}} - \mathsf{V}_{\mathsf{OH}}) \times \mathsf{I}_{\mathsf{OH}}),$ 

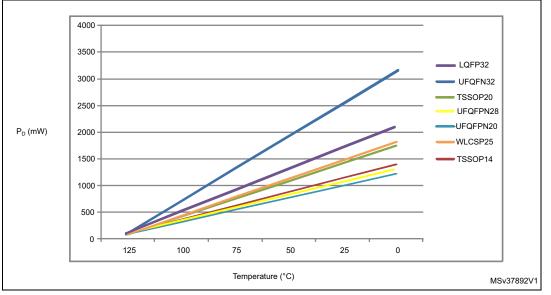
taking into account the actual V\_{OL} / I\_{OL} and V\_{OH} / I\_{OH} of the I/Os at low and high level in the application.



Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP32 - 7 x 7 mm / 0.8 mm pitch	60	
	Thermal resistance junction-ambient UFQFPN32 - 5 x 5 mm / 0.5 mm pitch	39	
	Thermal resistance junction-ambient WLCSP25 - 2.133 x 2.070 mm, 0.4 mm pitch	70	
$\Theta_{JA}$	<b>Thermal resistance junction-ambient</b> UFQFPN28 - 4 x 4 mm, 0.5 mm pitch	97	°C/W
	Thermal resistance junction-ambient UFQFPN20 - 3 x 3 mm, 0.5 mm pitch	102	
	Thermal resistance junction-ambient TSSOP20 - 169 mils	74	
	Thermal resistance junction-ambient TSSOP14 - 169 mils	95	

Table 75. Thermal characteristics





1. The above curves are valid for range 3.

#### 7.8.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org.



## 8 Ordering information

Table 76. STM32L011x3/4	ordering in	nform	ation	scher	ne			
Example:	STM32 L	011	К	4	Т	6	D	xxx
Device family								
STM32 = Arm-based 32-bit microcontroller								
Product type								
L = Low power								
Device subfamily								
011 = Access line								
Pin count								
K = 32 pins								
G = 28 pins								
E = 25 pins								
F = 20 pins								
D = 14 pins								
Flash memory size								
3 = 8 Kbytes								
4 = 16 Kbytes								
Package								
T = LQFP								
U = UFQFPN								
Y = WLCSP								
P = TSSOP								
Temperature range								
6 = Industrial temperature range, -40 to 85 °C								
7 = Industrial temperature range, -40 to 105 °C								
3 = Industrial temperature range, $-40$ to 125 °C								
Options								
No character = $V_{DD}$ range: 1.8 to 3.6 V and BO	R enabled							
D = $V_{DD}$ range: 1.65 to 3.6 V and BOR disabled	ł							
Packing								

TR = tape and reel

No character = tray or tube

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.



## 9 Revision history

Date	Revision	Changes
07-Dec-2015	1	Initial release.
11-Feb-2016	2	Features: modified current consumption in run mode, Cortex®-M0+ core frequency range and total number of timers.Updated ADC conversion consumption on cover page.Updated UFQFPN28 pinout: Figure 6: STM32L011x3/4 UFQFPN28 pinout and Table 13: Pin definitions.Updated Table 55: RAIN max for fADC = 16 MHz.Modified TS_CAL2 description in Table 57: Temperature sensor calibration values.
18-Mar-2016	3	<ul> <li>Changed minimum comparator supply voltage to 1.65 V on cover page.</li> <li>Added baudrate allowing to wake up the MCU from Stop mode in Section 3.15.3: Low-power universal asynchronous receiver transmitter (LPUART).</li> <li>Added number of fast and standard channels in Section 3.10: Analog-to-digital converter (ADC).</li> <li>Updated Table 16: Current characteristics to add the total output current for STM32L011GxUx.</li> <li>Changed V<sub>DDA</sub> minimum value to 1.65 V.in Table 18: General operating conditions.</li> <li>Updated Table 26: Current consumption in Sleep mode, Table 27: Current consumption in Low-power Run mode, Table 28: Current consumption in Low-power Sleep mode and Table 30: Typical and maximum current consumptions in Standby mode.</li> <li>Section 6.3.15: 12-bit ADC characteristics:</li> <li>Table 54: ADC characteristics:</li> <li>Distinction made between V<sub>DDA</sub> for fast and standard channels; added note 1.</li> <li>Updated t<sub>S</sub> and t<sub>CONV</sub>.</li> <li>Updated t<sub>S</sub> and t<sub>CONV</sub>.</li> <li>Updated Table 55: RAIN max for fADC = 16 MHz for f<sub>ADC</sub> = 16 MHz and distinction made between fast and standard standard channels.</li> <li>Updated measurement condition in Table 56: ADC accuracy.</li> </ul>

#### Table 77. Document revision history



Date	Revision	Changes
		Updated:
20-Jun-2016		<ul> <li>Features in cover page: Stop mode values, channels' number of DMA controller, I/Os' number, number of peripherals communication interface.</li> </ul>
	4	<ul> <li>Table 27: Current consumption in Low-power Run mode, Table 28: Current consumption in Low-power Sleep mode, Table 34: Low-power mode wakeup timings, Table 36: Low-speed external user clock characteristics</li> </ul>
12-Sep-2017	5	<ul> <li>Removed I/O operation from Table 3: Functionalities depending on the operating power supply range.</li> <li>Updated Section 3.4.4: Boot modes and added Note 7. in Table 13: Pin definitions.</li> <li>Changed USARTx_RTS and LPUARTx_RTS into USARTx_RTS_DE and LPUARTx_RTS_DE, respectively in Section 4: Pin descriptions.</li> <li>In Section 5: Memory mapping, replaced memory mapping schematic by reference to the reference manual.</li> <li>Updated introduction text in Section 6.2: Absolute maximum ratings to mention device mission profile and extended mission profiles.</li> <li>Added note in Table 49: I/O current injection susceptibility.</li> <li>Updated minimum and maximum values of I/O weak pull-up equivalent resistor (R<sub>PU</sub>) and weak pull-down equivalent resistor (R<sub>PD</sub>) in Table 50: I/O static characteristics.</li> <li>Updated minimum and maximum values of NRST weak pull-up equivalent resistor (R<sub>PU</sub>) in Table 53: NRST pin characteristics. Added note 2 related to the position of the external capacitor below Figure 26: Recommended NRST pin protection.</li> <li>Updated Section : I2C interface characteristics.</li> <li>Removed section USART/LPUART characteristics.</li> <li>Updated Figure 29: SPI timing diagram - slave mode and CPHA = 1(1) and Figure 31: SPI timing diagram - slave mode and CPHA = 1(1).</li> </ul>

Table 77. Document revision history



Date	Revision	Changes
12-Sep-2017	5 (continued)	<ul> <li>In Section 7: Package information:</li> <li>Added paragraph related to optional marking or inset/upset marks in all device marking sections.</li> <li>Updated Table 67: LQFP32 - 32-pin, 7 x 7 mm, 32-pin low-profile quad flat package mechanical data.</li> <li>Updated Figure 35: UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package outline and Table 68: UFQFPN32 - 32-pin, 5x5 mm, 0.5 mm pitch ultra thin fine pitch quad flat package mechanical data.</li> <li>Updated Figure 44: UFQFPN20 - 20-lead, 3x3 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package outline and Table 72: UFQFPN20 - 20-lead, 3x3 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package mechanical data.</li> <li>Added notes related to D and E1 in Table 73: TSSOP20 - 20-lead thin shrink small outline, 6.5 x 4.4 mm, 0.65 mm pitch, package mechanical data.</li> <li>Updated Figure 50: TSSOP14 - 14-lead thin shrink small outline, 6.5 x 4.4 mm, 0.65 mm pitch, package mechanical data.</li> <li>Section 8 renamed into Ordering information.</li> </ul>

Table 77. Document revision history



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